

1 **A high-performance, flexible, and cost-efficient auditory evoked response**  
2 **recording system appropriate for research purposes**

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21 **Abstract:**

22 The recording of auditory evoked responses (AER) is used in hospitals and  
23 clinics worldwide for hearing impairments detection and threshold estimation,  
24 and in research centers to understand and model the mechanisms involved in  
25 the process of hearing. This paper describes a high-performance, flexible, and  
26 inexpensive AER recording system. A full description of the hardware and  
27 software modules that compose the AER recording system is provided in this  
28 article. The performance of this system is evaluated by five experiments with  
29 both real and artificially synthesized auditory brainstem response (ABR) and  
30 middle latency response (MLR) signals at different intensity levels and  
31 stimulation rates. The results of this study point out that the flexibility of the  
32 described system is appropriate to record AER signals in several recording  
33 conditions. The AER recording system described in this article gives users a full  
34 control of the parameter settings involved in the AER recording process,  
35 incorporates a platform through which users are allowed to implement  
36 advanced signal processing methods, and its manufacturing cost is significantly  
37 lower than other current commercial alternatives. These advantages could be  
38 suitable in many research applications in the field of Audiology.

39 **Keywords:** Auditory evoked responses (AER); auditory brainstem response  
40 (ABR); middle latency response (MLR); evoked potentials; biomedical amplifier.

41

42 **List of abbreviations:**

43 AER: auditory evoked response; ABR: auditory brainstem response; MLR:  
44 middle latency response; ECochG: electrocochleography; SNR: signal-to-noise  
45 ratio; CMRR: common mode rejection ratio; EEG: electroencephalogram;  
46 AD/DA: analog to digital / digital to analog; ISI: interstimulus interval; CONV:  
47 conventional; MLS: maximum length sequences; CLAD: continuous loop  
48 averaging deconvolution; QSD: quasiperiodic sequence deconvolution; LS:  
49 least-squares; RSA: randomized stimulation and averaging; dB: decibel; I/O:  
50 input / output; nHL: normal hearing level; SPL: sound pressure level; ECochG:  
51 electrocochleography; ERP: event related potential.

52 **Text body:**

53 **1. INTRODUCTION**

54 The auditory evoked response (AER) is the electrical activity of the nervous  
55 system in response to a stimulus. This electrical activity is characterized by a  
56 number of voltage peaks of very low amplitude, called evoked potentials, which  
57 are generated in different parts of the auditory pathway. These evoked  
58 potentials can be classified according to their generator site and the time  
59 between the stimulus onset and the occurrence of the peaks (peak latency),  
60 which ranges between 1 ms and 0.5 second. The recording of the AER has  
61 been extensively used in human and animal studies for both clinical and  
62 research purposes due to its noninvasive nature. The auditory brainstem  
63 response (ABR) and the middle latency response (MLR) are AERs generated in  
64 the brainstem and in the auditory cortex respectively [7]. The ABR comprises a  
65 number of waves that occur during the first 10 ms from stimulus onset. These

66 ABR waves are identified by sequential roman numerals as originally proposed  
67 by Jewett and Williston [17]. Although up to seven peaks can be seen in the  
68 ABR, the most robust waves are I, III, and V. The MLR have latencies from 10  
69 to 60 ms and comprise the components  $N_a$ ,  $P_a$ ,  $N_b$ ,  $P_b$ . The longer component of  
70 the MLR is usually affected by attention and is difficult to record under sedation.  
71 The recording of these signals is commonly used in hospital and clinics  
72 worldwide as a hearing screening tool, to detect the hearing threshold, and  
73 hearing impairments such as vestibular schwannoma and Ménière's disease.  
74 Furthermore, the analysis of the AER may help understand the underlying  
75 mechanisms of the process of hearing [20,23,34]. The recording process of  
76 these signals involves the setting-up of a wide range of factors [26].

77 This paper describes in detail a high-performance, flexible, and inexpensive  
78 AER recording system. Although there already exist several clinical systems  
79 that allow the recording of the AER, most of them are expensive, the control  
80 over most of the parameter settings is limited, and give no access to raw  
81 recorded data [1]. In contrast, the AER recording system described in this article  
82 gives users a full control of the parameter settings. Users are able to set the  
83 intensity level of stimulation, select the number of auditory responses for  
84 average, use the conventional method of stimulation or any other more  
85 advanced technique, set the stimulation frequency, select the analog-to-digital  
86 sampling frequency, choose the order and band-pass cut-off frequencies for  
87 digital filters, select the polarity of stimulation and nature of the stimuli (clicks,  
88 chirps, tone pips, etc.), or implement advanced artifact rejection techniques. In  
89 addition, this system gives access to raw recording data, thus advanced signal  
90 processing methods can be implemented offline. The performance of this



91 system is evaluated through five experiments with both real and artificially  
92 synthesized ABR and MLR signals recorded at different intensity levels and  
93 stimulation rates. The flexibility, along with the high-performance and cost-  
94 efficient nature of the AER recording system described in this article, might be  
95 suitable to carry out research activities of different nature in the field of  
96 Audiology.

## 97 **2. SYSTEM ARCHITECTURE**

### 98 **2.1. System overview**

99 The procedure for recording the AER is schemed in figure 1. This process  
100 includes the presentation of auditory stimuli and the recording of their  
101 corresponding electrical response (sweep) by surface electrodes. A high  
102 amplification of this signal is required due to the low amplitude of the AER  
103 (usually less than 1  $\mu$ V). The recorded signal is usually highly contaminated by  
104 different types of artifacts, such as myogenic noise related to the muscular  
105 activity of the subject, electrical noise derived from the amplifier,  
106 electromagnetic and radiofrequency interferences, etc. The conventional  
107 method used to reduce the effects of these artifacts and improve the signal-to-  
108 noise ratio (SNR) of the response is the average of a large number of sweeps  
109 whose corresponding stimuli are periodically presented [4,8,37]. This system is  
110 battery powered in order to reduce the artifact generated by the electric power  
111 network. The stimulation of the auditory system is conventionally performed by  
112 0.1 ms duration clicks in rarefaction polarity in order to evoke a synchronous  
113 firing of a large number of neurons, however, this system allows the  
114 implementation of other stimulus types such as tone burst, filtered clicks, chirps,

115 noise stimuli, and speech stimuli [13]. The intensity level can be controlled by  
116 setting the amplitude of the stimulation signal. A signal composed of a burst of  
117 stimuli is generated by the laptop for both stimulation and synchronization  
118 purposes. This signal is sent synchronously by the left and right outputs of an  
119 analog-to-digital / digital-to-analog (AD/DA) soundcard. The right output is  
120 connected to the left input for the synchronization of the stimuli. The left output  
121 is connected to a pair of insert earphones, through which the stimulation signal  
122 excites the auditory system of the subject, thus generating the AER. This  
123 biological signal, plus noise, is recorded by three electrodes placed on the skin  
124 at different positions on the head. The electroencephalogram (EEG) recorded  
125 by the electrodes is amplified and band-pass filtered. The auditory response  
126 after filtering and amplification is recorded synchronously along with the  
127 synchronization signal by the right and left inputs of the external AD/DA  
128 soundcard. The software routines of this system implement the digital signal  
129 processing methods necessary to obtain the AER. Figure 2 shows a picture of  
130 the electronics of the amplifier (left) and the hardware elements that compose  
131 the AER recording system (right). Table 1 presents a rough cost list of the  
132 elements that include the AER recording system. This table was built  
133 considering the price list of a well-known international electronics supplier. The  
134 cost analysis presented in this table shows that the cost of the elements and  
135 materials involved in the AER recording system prototype described in this  
136 paper (laptop not included) is around 950 USD.

137

## 138 **2.2. Hardware specifications**

### 139 **2.2.1. Amplifier**

140 The amplifier is composed of four stages: preamplification, band-pass filtering,  
141 amplification, and active ground circuitry. The electronic schematic of the  
142 amplifier is shown in figure 3. The stage of preamplification provides a moderate  
143 gain in order to avoid saturation in later stages. This stage is based on the  
144 instrumental amplifier INA128 (Texas Instruments Inc., Dallas, TX). This  
145 differential amplifier was chosen because of its high common mode rejection  
146 ratio (CMRR), low power, low noise ( $8\text{ nV}/\sqrt{\text{Hz}}$ ), and easy control of the gain.  
147 The band-pass filtering stage removes the frequencies out of the scope of the  
148 AER, amplifying only the band of interest. This stage comprises four second  
149 order Sallen-Key filters (2 x high pass & 2 x low pass). The values of the  
150 resistors and capacitances that implement the filtering stage define the  
151 bandwidth of the amplifier. The bandwidth of the amplifier must be selected  
152 considering the characteristic frequencies of each AER. Table 2 shows the  
153 characteristic bandwidth for recording ABR and MLR signals, along with  
154 suggested values of resistors and capacitances that implement the high pass  
155 and low pass filtering stages of the amplifier. These analog filters insert a phase  
156 distortion on the recorded signal that must be adjusted by software. This phase  
157 shift is  $560\ \mu\text{s}$  for the ABR amplifier and  $80\ \mu\text{s}$  for the MLR amplifier. The  
158 amplification stage after filtering sets the required level of amplitude on the EEG  
159 to be recorded by the analog to digital converter. The active ground circuit is  
160 designed to reduce the common mode voltage of the recorded signal. The  
161 electric field generated by the electric network can induce a common mode  
162 voltage on the subject. This common mode voltage is amplified, inverted, and

163 inserted back to the subject by the active ground circuit, thus reducing  
164 significantly the common mode voltage on the subject. The operational  
165 amplifiers OPA227 (Texas Instruments Inc., Dallas, TX) used in this circuit were  
166 chosen because of their very low noise voltage ( $3\text{ nV} / \sqrt{\text{Hz}}$ ), high CMRR (130  
167 dB), and high precision. The Bode diagrams on Figure 4 show the bandwidth  
168 and the phase shift of the amplifiers for ABR and MLR signals. The gain of the  
169 amplifier reaches the value  $G_A = 20.000$  (86 dB) for the band-pass frequencies,  
170 with a filter slope of 24 dB/oct. Figure 5 represents a linearity analysis for the  
171 ABR amplifier. This figure represents a 10 ms sinusoidal signal inserted on the  
172 amplifier (input signal) versus its corresponding output signal. The slope of this  
173 curve represents the gain of the amplifier (86 dB). The amplitude of the input  
174 signal was chosen to obtain a slightly saturated output signal. The frequency of  
175 the input signal was set on 1087 Hz to obtain an output signal with phase  
176 distortion zero. This analysis suggests that the behavior of the amplifier is  
177 especially lineal when the input signal is in the range  $[-0.3 +0.3]$  mV, a common  
178 situation considering that the recorded EEG does not usually exceed  $50\ \mu\text{V}$   
179 [13]. Thus, the dynamic range of the amplifier is  $600\ \mu\text{V}$ . The consumption of  
180 this circuit is 28.2 mA, which gives the device an operating time of more than 6  
181 hours for standard rechargeable 9V batteries. The safety of the subject under  
182 exploration is granted, on one hand, by the battery powered nature of the  
183 system, which prevents any possible electrical shock derived from the electrical  
184 network; and on the other hand, by the  $1\ \text{M}\Omega$  resistor that connects the active  
185 ground electrode to the subject, which limits the leakage current introduced to  
186 the subject to  $9\ \mu\text{A}$ , meeting the electrical safety requirements of the  
187 international standard IEC 60601-1 [38].

188        **2.2.2.    Electrodes**

189    Electrodes transform ionic currents (mechanism of conduction of bioelectrical  
190    signals on tissues) into electrical currents that conduct the evoked potentials  
191    from the subject to the recording system. Since the electrodes are the first  
192    components on signal recording, the noise level generated at them should be  
193    minimized. The preferable electrodes to reduce contact potential, typically used  
194    in AER recording, are silver coated with silver chloride (Ag/AgCl) surface  
195    electrodes, composed of a silver conductor (electrode) immersed into a silver  
196    chloride salt dissolution (electrolyte). Electrolytic paste is used as a mean of  
197    union between the electrode and the skin in order to reduce the contact  
198    electrode impedance. The contact impedance of the junction between the scalp  
199    and the electrodes should be kept as low as possible to minimize the magnitude  
200    of induced electromagnetic artifacts and to reduce the capacitive coupling  
201    effects of the electrode cables and external power lines [13,26]. This contact  
202    impedance can be reduced by a softly scrape of the skin with alcohol or other  
203    cleansing agent. The electrode-skin contact impedance can be measured either  
204    by any commercial alternating-current impedance meter, or by implementing the  
205    circuit diagram of any impedance meter described in the literature, e.g., [11,12].  
206    Impedances lower than 5 k $\Omega$  at the working frequencies can be considered  
207    acceptable. The electrodes impedance should be balanced to avoid common  
208    mode artifacts. The placement of the electrodes can be done in accordance  
209    with the standard positions defined by the International 10-20 and 10-10  
210    Systems [16,19]. Active, ground, and reference electrodes can be situated at  
211    the high forehead (Fz), low forehead (Fpz), and ipsilateral mastoid (TP9/TP10)  
212    respectively, as shown in figure 1. Active and reference electrodes are

213 connected to the differential inputs of the amplifier. The ground electrode  
214 connects the active ground input of the amplifier.

### 215 **2.2.3. Analog-to-Digital conversion**

216 The analog-to-digital conversion is carried out by an external soundcard  
217 connected to the laptop through the USB port. This device presents the  
218 advantages of simplicity and a better performance than most of soundcards  
219 integrated on laptops. Table 3 shows a summary of the features of the AD/DA  
220 soundcard. The number of bits of quantization and the sampling rate can be  
221 controlled by the user.

222 The amplitude precision of the analog-to digital conversion is determined by the  
223 number of bits of quantization. Considering the recording of ABR signals, the  
224 analog-to-digital converter should be able to measure on the range 2 nV (10%  
225 precision of a standard 20 nV amplitude of a wave II) to 200  $\mu$ V (highest  
226 expected recorded level of an EEG). This is a ratio of 100.000, and corresponds  
227 to a dynamic range of 100 dB. Considering that an AD/DA of n bits has a  
228 dynamic range of  $6 \cdot n$  dB, the required number of bits of the AD/DA to be able to  
229 record ABR with a precision of a 10% is about 16 bits. In addition to this, the  
230 process of sweeps averaging increases the precision of the measure, reducing  
231 the quantization noise [26]. Therefore, the use 16 bits of quantization is enough  
232 to record AER with sufficient precision.

233 The sampling rate must be greater than the double of the highest frequency  
234 component present in the signal in order to prevent aliasing [21]. However, the  
235 low-pass filters of the filtering stage in the amplifier just attenuate (not eliminate)  
236 the frequency components greater than the cutoff frequency ( $f_c$ ). The aliasing

237 errors from all frequency components would be prevented only when the  
238 sampling rate is set to twice the frequency at which the filter attenuates the  
239 signal by more than the dynamic range of the AD/DA. Considering a standard  
240 AD/DA converter, the frequency at which this occur is  $f' = f_c \cdot 2^{(D-3)/S}$ , where  $f_c$   
241 is the cutoff frequency, D is the dynamic range of the AD/DA in dB, and S is the  
242 slope in dB per octave [26]. Therefore, to avoid even 1-bit aliasing errors, the  
243 sampling rate ( $f_s$ ) must be  $f_s = 2 \cdot f_c \cdot 2^{(D-3)/S}$ . Since the AER recording system  
244 described in this article includes an anti-aliasing filter with a cutoff frequency of  
245 3000 Hz and a steep slope of 24 dB per octave used in conjunction with a 16-bit  
246 AD/DA, the sampling rate must be over 22982 samples per second to avoid all  
247 aliasing errors. Hence, a sampling rate of 25 kHz could be appropriate to avoid  
248 all aliasing errors and at the same time, prevent undesired effects of  
249 oversampling.

#### 250 **2.2.4. Transducer**

251 Earphones provide the stimulation of the auditory system of the subjects by  
252 transducing the electrical energy of the stimulation signal into acoustical energy  
253 (sound). The tubal insert earphones Etymotic ER3A (Etymotic Research, Inc.,  
254 Elk Grove Village, IL) were chosen for this application because of their flat  
255 response on a wide band of frequencies, their isolation from external noise, and  
256 their fast response to typical click stimuli, which enables a synchronous firing of  
257 inner hair cells [13]. Other standard earphones such as the Telephonics TDH-  
258 39, -49, -50 (Cadwell Laboratories, Inc., Kennewick, WA) could also be used.

## 259 **2.3. Software specifications**

260 The software modules involved in the AERs recording process are presented in  
261 figure 6. The first step on data acquisition is the generation of the stimulation  
262 signal. The conventional stimulation technique consists of the presentation of  
263 stimuli with a constant inter-stimulus interval (ISI) greater than the averaging  
264 window to avoid overlapping responses [4]. Other more advanced methods can  
265 also be implemented to obtain AER at high stimulation rates such as maximum  
266 length sequences (MLS) [9], continuous loop averaging deconvolution (CLAD)  
267 [5,22], quasiperiodic sequence deconvolution (QSD) [18], least-squares  
268 deconvolution (LS) [2,3], and randomized stimulation and averaging (RSA) [27].  
269 The stimulation of the auditory system is typically performed by 0.1 ms duration  
270 clicks in rarefaction polarity in order to evoke a synchronous firing of a large  
271 number of neurons [13]. Other types of stimuli can also be implemented such as  
272 tone bursts, filtered clicks, paired clicks, plops, chirps, modulated tones,  
273 stimulus trains, noise stimuli, and speech stimuli. The parameters type of  
274 stimuli, intensity level, clicks duration, clicks polarity, stimulation rate, and  
275 number of recorded sweeps can be controlled in this module. The “Stimulation  
276 & Recording” module consists of (a) the synchronous reproduction of the  
277 stimulation signal and (b) the synchronous recording of the stimulation signal  
278 and the digitized electroencephalogram (EEG). The user has the control of the  
279 number of quantization bits and the sampling rate on this step. The function of  
280 the “Scaling” module is to convert the recorded signal ( $A_x$ ) into its  
281 corresponding value in microvolts at the electrodes. Considering  $G_A$  the gain of  
282 the amplifier for the band-pass frequencies and  $G_S$  the gain of the AD/DA, the  
283 scaled value in microvolts at the electrodes is  $A_{scaled}(\mu V) = A_x \cdot \frac{1}{G_S} \cdot \frac{1}{G_A} \cdot 10^6$ .



284 The values of  $G_A$  and  $G_S$  are estimated on the calibration process, which is  
285 described in section 2.4.1. The “AER enhancement” module incorporates  
286 algorithms to increase the quality of the response such as digital filtering and  
287 artifact rejection techniques. The “Synchronization” module uses the recording  
288 of the stimulation signal as trigger to determine the samples at which stimuli  
289 occurs. The “AER calculation” module runs the necessary algorithms to obtain  
290 the AER according to the method used in the stimulation process. This software  
291 module also compensates the phase distortion inserted by the analog filters of  
292 the amplifier on the recorded AER. Finally, the “Storage” module saves the raw  
293 data, the processed variables, and other important parameters into a file on the  
294 database. The parameters involved in the process of recording AERs can be  
295 managed from a graphical user interface (GUI). The structure of this multimedia  
296 platform can be designed according to the specific requirements of the users.  
297 Figure 7 shows an example of an interactive front-end of the AER recording  
298 system, in which the user has the control of recording parameters such as the  
299 interstimulus interval of the stimulation sequence (ISI), the number of recorded  
300 sweeps, the intensity level and the duration of the click. This platform also  
301 allows the use of specific signal processing techniques to obtain signals of  
302 higher quality such as digital filtering, frame rejection, and digital blanking.  
303 Additional information such as the number of accepted and rejected frames, the  
304 acceptance ratio, and the recorded EEG are also provided. In this example of  
305 front-end, the auditory evoked response is shown in a graph, as well as a  
306 history of previous recorded signals. An example of software routine that  
307 implements the recording of AER using the conventional method is available in  
308 MATLAB (The Mathworks, Inc., Natick, MA) code as supplementary material  
309 (Additional file).

## 310 **2.4. Calibration**

### 311 **2.4.1. Calibration of $G_A$ and $G_S$**

312 The calibration process consists of estimating the values of the gain of the  
313 amplifier for the band-pass frequencies ( $G_A$ ) and the gain of the AD/DA ( $G_S$ ) to  
314 perform a correct scaling of the recording signal. The value of  $G_A$  can be  
315 estimated directly from the Bode diagram of the amplifier. The value of  $G_S$  is  
316 related to the intensity level of the input line of the AD/DA soundcard. This  
317 parameter can be configured from the audio settings of the laptop. Medium  
318 intensity level is recommended to avoid possible nonlinearities. The value of  $G_S$   
319 can be estimated by correlating a recorded a signal whose maximum amplitude  
320 in volts is known ( $V_{hi}$ ) with its corresponding value of the recorded signal ( $X_{hi}$ ),  
321  $G_S = V_{hi} / X_{hi}$ .

### 322 **2.4.2. Calibration of the intensity level**

323 The calibration of the intensity level consists of the measure of the stimulus  
324 magnitude, necessary for providing an accurate and uniform evaluation of the  
325 evoked responses. The standard audiometric calibration methods include dB  
326 normal hearing level (nHL) and dB sound pressure level (SPL) [6]. The intensity  
327 level 0 dB nHL represents the hearing threshold for normal hearing subjects.  
328 This intensity level can be established as the mean value of the intensity level at  
329 which stimuli are just detectable in a set of 15 to 20 subjects with no auditory  
330 dysfunction (normal hearing subjects) [13]. The intensity level of a stimulus in  
331 terms of dB SPL is estimated as  $20 \cdot \log_{10} \left( \frac{P_x}{P_{ref}} \right)$ , being  $P_x$  the pressure of the  
332 stimulus and  $P_{ref}$  the reference pressure, whose typical reference value is 20

333  $\mu\text{Pa}$ . A complete description of the procedure to calibrate the reference zero is  
334 described by the international standard ISO 389 [15,25]. In this system, the  
335 calibration of the stimuli is performed according to the aforementioned  
336 international standard. The intensity level can be controlled by the user through  
337 the output voltage of the stimulation signal. Given  $V_{\text{ref}}$  as the amplitude voltage  
338 of a stimulation signal that presents an intensity level of 0 dB nHL, the  
339 amplitude voltage necessary to present an intensity level of X dB nHL can be  
340 obtained according to:  $V_X = V_{\text{ref}} \cdot 10^{X/20}$ .

## 341 **2.5. Scalability**

342 The use of multiple-channel systems might be required in certain research  
343 applications, e.g., the use of binaural stimulation for simultaneous screening in  
344 both ears, the use of contralateral masking to ensure monaural stimulation, and  
345 the simultaneous screening of ABR and electrocochleography (ECoChG) [24].  
346 The AER recording system described in this system is scalable. A multichannel  
347 version of this system can be set up using an AD/DA converter of multiple  
348 channels, and multiple units of the amplifier. Considering that the price of a  
349 standard 4 channels AD/DA sound card is about 150 USD, and that the rough  
350 manufacturing cost of an amplifier unit is about 200 USD, the implementation of  
351 a 4 channels AER recording system would reach a total manufacturing cost of  
352 about 1250 USD.

353

## **3. ASSESSMENT**

354 The performance of the AER recording system described in this article is  
355 evaluated by a number of experiments conducted on one normal hearing

356 subject (#S1: male, 28 yr). The subject explored in these experiments was  
357 informed about the experimental procedure and possible side effects of the test,  
358 and gave consent for the use of the data. The calibration of the intensity level  
359 was carried out according to the international standard ISO 389 [15,25]. The  
360 equivalent 0 dB nHL corresponds to 36.4 dB SPL. The recording procedure of  
361 these experiments was approved by the Clinical Research Ethics Committee of  
362 the San Cecilio University Hospital and by the Human Research Ethics  
363 Committee of the University of Granada (Reference No. 826014263-14263-4-9),  
364 in accordance with the Code of Ethics of the World Medical Association  
365 (Declaration of Helsinki) for experiments involving humans. Additionally, this  
366 section introduces an outline of related research activities carried out by the  
367 AER recording system described in this paper.

368 Experiment 1 was driven to simulate the recording of ABR and MLR signals and  
369 assess the performance of the AER recording system. The ABR and MLR  
370 signals used in this experiment (original pseudopotentials) were obtained from  
371 #S1 using 10.000 click stimuli presented at a rate of 33 Hz for ABR and 3.3 Hz  
372 for MLR at an intensity level of 70 dB nHL. A burst of 10.000 pseudopotentials  
373 was digitally synthesized for each type of signal. The amplitude of both signals  
374 was reduced by a voltage divider to obtain signals of 0.2  $\mu$ V for ABR and 0.5  $\mu$ V  
375 for MLR. The burst of low-amplitude pseudopotentials was amplified, recorded  
376 by the AD/DA soundcard, and digitally processed according to the recording  
377 procedure described in section II. Figure 8 shows the original and recorded  
378 pseudopotentials for ABR and MLR signals. The most important components of  
379 these signals are marked on the figure. This figure shows that the AER  
380 recording system described in this article can be used to obtain signals similar

381 in morphology and amplitude to ABR and MLR since the major components of  
382 these signals can be easily identified, remain on the same latency, and present  
383 similar amplitude.

384 Experiment 2 analyzes the effects of noise reduction through sweeps  
385 averaging. Figure 9 shows ABR and MLR signals obtained from #S1 at a  
386 different number of averaged sweeps. The stimuli used on this experiment were  
387 clicks presented at 70 dB nHL at a stimulation rate of 33 Hz for ABR and 8 Hz  
388 for MLR. This figure shows that the quality of the AER increases with the  
389 number of averaged sweeps. The main waves of these signals start to be  
390 identified with at least 500 sweeps. The recordings obtained with 20.000  
391 sweeps are of a high quality but require a long test time, especially for MLR  
392 signals. A number of 2000 sweeps can be found appropriate to reach a  
393 compromise between recording time and quality of the recordings. However, the  
394 recording of AER obtained with larger number of averaged sweeps can be  
395 interesting in certain applications, such as the study of neural adaptation, that  
396 require the analysis of high quality AER and do not impose significant  
397 restrictions on the recording test time [34].

398 Experiment 3 evaluates the influence of intensity level on the morphology of  
399 ABR signals. Figure 10 shows ABR signals from #S1 obtained at intensity levels  
400 of stimulation that vary from 5 to 80 dB nHL, in steps of 5 dB. 5.000 sweeps  
401 were recorded for each ABR signal. Waves I, III, and V are labeled on the ABR  
402 signal obtained at 80 dB nHL. This experiment shows that the amplitude of the  
403 most relevant waves decreases and their corresponding latency increases as  
404 the stimulation intensity level decreases. Wave V remains as the most robust

405 component, that in this experiment can be clearly identified up to 15 dB nHL.  
406 These results are in accordance with previous literature [14,17].

407 Experiments 4 and 5 analyze the effects of stimulation rate on the morphology  
408 of ABR and MLR signals respectively. Figure 11 shows ABR signals from #S1  
409 obtained at stimulation rates up to to 250 Hz using the randomized stimulation  
410 and averaging (RSA) [27], the quasiperiodic sequence deconvolution (QSD)  
411 [18], and the conventional (CONV) techniques [4]. All recordings were obtained  
412 using 5.000 averaged sweeps stimulated with clicks at 70 dB nHL. The amount  
413 of jitter used in the stimulation sequences for RSA and QSD was 4 ms. The  
414 jitter of a stimulation sequence measures the grade of dispersion of the ISI  
415 compared to a periodical presentation of the stimuli, i.e., the ISI of stimuli  
416 presented at a rate of 25 Hz with a jitter of 4 ms would vary between 38 and 42  
417 ms.

418 Both RSA and QSD techniques are valid methods to obtain ABR signals at very  
419 high stimulation rates (greater than 100 Hz). Waves I, III, and V can be clearly  
420 identified in all recordings, although the ABR signal obtained with QSD at 250  
421 Hz is slightly noisier. This figure shows the normal changes on the morphology  
422 of the ABR as stimulation rate increases: amplitude of waves decrease and  
423 latencies increase, with a deeper shift on the most central waves. Figure 12  
424 shows MLR signals from #S1 obtained at stimulation rates from 8 to 125 Hz  
425 obtained with the RSA technique with a jitter of 16 ms, using click stimuli  
426 presented at 70 dB nHL. The V, Na, Pa, Nb, Pb components can be identified at  
427 all stimulation rates. These components are labeled on the MLR signal obtained  
428 at 125 Hz. The MLR signal obtained at 40 Hz presents a resonance, in which  
429 the Na, Pa, Nb, and Pb components are in phase (occur at the same time

430 relative to the stimulus) and become superimposed. This phenomenon is  
431 generally known as 40-Hz event-related potential (ERP) and was first described  
432 by Galambos et al. (1981) [10]. The 40-Hz ERP presents advantages for the  
433 estimation of the auditory threshold due to its large amplitude (usually greater  
434 than 1  $\mu$ V).

435 In addition to these five experiments, the AER recording system described in  
436 this paper has been successfully used in related research activities. This  
437 system was used to develop (a) the RSA method, a technique that allows the  
438 recording of AER at high rates [27]; the separated response method, which  
439 allowed for the first time the study of the fast and slow mechanisms of  
440 adaptation in humans [29,34]; (c) the fitted parametric peaks (FPP) method,  
441 which provides an automatic evaluation of the quality of ABR signals and a  
442 parameterization of the most important waves in terms of amplitude, latency  
443 and width [30,35]; (e) studies to test whether or not high stimulation rates could  
444 save recording time [28,36]; (f) an automatic auditory response detection  
445 paradigm based on response tracking [31]; (g) a study of the effects of  
446 averaging and deconvolution in ABR and MLR signals using the RSA method  
447 [32]; and (h) a deconvolution method based on randomized stimulation using  
448 artifact rejection methods in the frequency domain [33].

449

#### **4. DISCUSSION**

450 This paper provides a full description of a flexible, high-performance, and  
451 inexpensive auditory evoked response recording system. The system described  
452 in this article includes an amplifier, an external sound card that acts as an  
453 AD/DA converter of two I/O channels, electrodes, cables and connectors, and a

454 laptop with software modules that run the algorithms for the stimulation  
455 sequence generation, the production of the stimuli and the recording of the  
456 sweeps, the scaling of the recorded EEG, the synchronization of the sweeps  
457 with their associated stimuli, the processing of data according to the specific  
458 stimulation method to obtain the AER (conventional, MLS, QSD, CLAD, LS,  
459 RSA, etc.), and finally, the storage of the EEG and the AER into a file.

460 The open nature of this system provides the flexibility required on many  
461 research applications. Almost every parameter involved in the AER recording  
462 process could be defined and controlled. For instance, this system gives the  
463 user the control on parameters such as the nature, duration and polarity of  
464 stimuli, the number of averaged sweeps, the intensity level, and the stimulation  
465 rate. The software platform of this system allows the implementation of  
466 advanced stimulation methods, such as RSA and QSD, which allows the  
467 recording of AER signals at high rates of stimulation, digital filtering to enhance  
468 the quality of the recordings, and the use of artifact rejection methods. In  
469 addition, the recording of the raw electroencephalogram may be of interest to  
470 implement advanced signal processing methods offline. Furthermore, the  
471 scalability of the system allows the implementation of a multiple-channel design,  
472 which might be useful in certain research applications such as the use of  
473 binaural stimulation for simultaneous screening in both ears, the use of  
474 contralateral masking to ensure monaural stimulation, and the simultaneous  
475 screening of ABR and electrocochleography (ECoChG) [24].

476 The performance of this system was evaluated through a number of  
477 experiments that include (a) the recording of artificially synthesized ABR and  
478 MLR signals (pseudopotentials), (b) the recording of real ABR and MLR signals



479 of different quality using a varying number of averaged sweeps, (c) the analysis  
480 of the influence of the intensity level on the morphology of ABR signals, (d) and  
481 the study of the effects of stimulation rate on the morphology of ABR and MLR  
482 signals. Some of the results obtained in these experiments are especially  
483 remarkable, such as the ABR signal obtained at 250 Hz and the MLR signal  
484 recorded at 125 Hz (experiments 4 and 5). In addition to these experiments, the  
485 AER recording system proposed in this article has been proven to be effective  
486 in several preceding studies, e.g., this architecture was used (a) to develop the  
487 RSA method and compare its performance with the QSD technique through  
488 ABR signals recorded from 8 subjects at different stimulation rates [27]; (b) to  
489 do a study of the fast and slow mechanisms of adaptation in humans by  
490 analyzing the morphology of ABR signals obtained with the separated  
491 responses methodology [29,34]; (c) to develop and evaluate different  
492 approaches of automatic quality assessment and response detection methods  
493 [30,31,35]; (d) to carry out a study to test whether or not high stimulation rates  
494 could save recording time [28,36]; (e) an analysis of the effects of adaptation  
495 and deconvolution of ABR and MLR signals with RSA [32]; and (f) to develop a  
496 method that allows the deconvolution of overlapping responses with  
497 randomized stimulation using frequency domain-based artifact rejection  
498 methods [33]. The analysis of the results of the experiments carried out in this  
499 article along with the results obtained in the aforementioned preceding studies  
500 [27-36], point out that the AER recording system described in this article can be  
501 efficiently used to record ABR and MLR signals in different recording conditions.

502 Despite already exist several clinical devices for recording AERs, most of them  
503 are expensive and suffer from a lack of flexibility since they are designed for

504 specific applications (e.g., hearing threshold estimation). The commercial  
505 systems designed for research applications are more flexible than the  
506 aforementioned clinical devices; however, the flexibility of these systems is  
507 limited by the performance of their associated software, and their acquisition  
508 price is usually high since it includes not only the cost of the materials, but also  
509 costs derived from marketing, distribution, technical support, profit margin, etc.  
510 In contrast, the rough cost for the implementation of a prototype of the  
511 described AER recording system including circuitry, connectors, box, external  
512 AD/DA soundcard, the Etymotic ER-3A insert earphones, electrodes, and  
513 cables (laptop not included) is less than 1000 USD. The cost-efficient nature of  
514 the auditory evoked response recording system described in this article, along  
515 with its high-performance and flexibility, could be valuable in several research  
516 applications in Audiology.

## 517 **5. CONCLUSION**

518 This article describes in detail the hardware and software elements of an  
519 auditory evoked response recording system. The performance of this system  
520 has been assessed by five experiments with both real and artificially  
521 synthesized ABR and MLR signals in different recording conditions. The high-  
522 performance, flexible, and cost-efficient nature of the AER recording system  
523 described in this article could be valuable in several research applications in the  
524 field of Audiology.

## 525 **ADDITIONAL FILE**

526 Additional file 1: Example of MATLAB routine that implements the recording of  
527 AER using the conventional method.

## COMPETING INTERESTS

528

529 There are no conflicts of interest associated with this research article. The  
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532

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#### 666 **Figure legends:**

- 667 • Figure 1. General scheme of the AER recording system.
- 668 • Figure 2. Picture of the electronics of the amplifier (left) and hardware  
669 modules of the AER recording system (right).
- 670 • Figure 3. Electronic circuit diagram of the amplifier.
- 671 • Figure 4. Bode diagram of the amplifier.
- 672 • Figure 5. Input signal versus output signal graph for a linearity analysis of  
673 the ABR amplifier.
- 674 • Figure 6. Software modules diagram.
- 675 • Figure 7. Interactive front-end of the AER recording system. This  
676 multimedia platform allows the user a full control of all parameters  
677 involved in the AER recording process.

- 678 • Figure 8. Recording of low-amplitude digitally synthesized signals similar  
679 in morphology to ABR and MLR potentials.
- 680 • Figure 9. Influence of the number of averaged sweeps on the quality of  
681 ABR and MLR signals.
- 682 • Figure 10. ABR signals obtained at different intensity levels of  
683 stimulation.
- 684 • Figure 11. ABR signals recorded at different stimulation rates using the  
685 randomized stimulation and averaging (RSA), the quasiperiodic  
686 sequence deconvolution (QSD), and the conventional (CONV) methods.
- 687 • Figure 12. MLR signals obtained at different stimulation rates using the  
688 randomized stimulation and averaging (RSA) technique.

689 **Table legends:**

- 690 • Table 1. Rough cost analysis of the elements that compose the AER  
691 recording system.
- 692 • Table 2. Frequency bandwidth of different AERs and suggested values of  
693 resistors and capacitances that implement the high pass and low pass  
694 filtering stages of the amplifier.
- 695 • Table 3. Features of the AD/DA soundcard.

696

697

698

## Supplementary Material

700 **Example of MATLAB routine that implements the recording of AER using the conventional**  
701 **method**

```

702 %% PARAMETERS INITIALIZATION
703 fs = 25e3; % Sampling rate
704 Name_File = 'EEG_Example'; % Name of the file
705 ER = 1; % Evoked response: ER=0 for ABR, ER=1 for
706 MLR
707 if(ER)
708     window = 12e-3; % Time window of 12 ms for ABR
709     Low_freq = 100; % Low-pass frequency for digital filter
710     High_freq = 3000; % High-pass frequency for digital filter
711     Phase_delay = 15; % Phase distortion compensation (560 us)
712 else
713     window = 100e-3; % Time window of 100 ms for MLR
714     Low_freq = 10; % Low-pass frequency for digital filter
715     High_freq = 3000; % High-pass frequency for digital filter
716     Phase_delay = 3; % Phase distortion compensation (80 us)
717 end
718 AER = zeros(window*fs,1); % AER initialization
719 ISI = 0.030; % Interstimulus interval of the sequence
720 in ms
721 N_Sweeps = 2000; % Number of recorded sweeps
722 Click_Duration = 120e-6; % Duration of the click in s
723 Ga = 1250; % Gain of the amplifier (calib)
724 Gs = 1.0461; % Gain of the AD/DA soundcard (calib)
725 Filter_Order = 4; % Order of the digital filters
726 V_ref = 9.8465e-5; % Absolute intensity level for 0 dBnHL
727 (calib)
728 I = 70; % Intensity level in dBnHL
729 clear ER window
730
731 %% STIMULATION SIGNAL GENERATION
732 x(1:Click_Duration*fs,1) = -1; % Pattern of the rarefaction
733 click
734 h(1:ISI*fs:N_Sweeps*ISI*fs) = 1; % h=1 -> start of the stimuli
735 Seq = conv(x,h); % Signal sequence generation
736 % Channel 1 - Stimulation signal. Channel 2 - Synchronization signal
737 Seq(:,2) = Seq(:,1); % 2-channels sequence
738 t_blocking = floor(length(Seq)/fs); % Recording test time
739 Seq(:,1) = Seq(:,1)*V_ref*10^(I/20); % Seq - intensity level
740 calibrated
741 clear Click_Duration N_Sweeps ISI x h V_ref I
742
743 %% STIMULATION & RECORDING
744 x = audioplayer(Seq,fs,16);
745 play(x);
746 sound(Seq,fs,16);
747 recorder = audiorecorder(fs,16,2);
748 recordblocking(recorder,t_blocking);
749 y = getaudiodata(recorder);
750 clear t_blocking Seq x recorder
751
752 %% SCALING

```

```

753 EEG = y(:,1)-mean(y(:,1));           % Remove the offset of the input
754 signal
755 EEG = EEG/Ga/Gs*1e6;                 % EEG calibrated in microvolts
756 Sinc = y(:,2)-mean(y(:,2));         % Remove the offset of the input
757 signal
758 clear y Ga Gs
759
760 %% AER ENHANCEMENT
761 [b a] = butter(Filter_Order,[Low_freq High_freq]*2/fs,'bandpass');
762 Resp = filter(b,a,EEG);              % EEG after digital filtering
763 clear a b Filter_Order Low_freq High_freq
764
765 %% SYNCHRONIZATION
766 % Sinc is replaced with samples of amplitude over the 70% of the
767 maximum
768 Sinc = find(Sinc>0.7*max(Sinc));
769 % Only the first sample is relevant. The following 10 samples are
770 removed.
771 m(1) = Sinc(1);                      % m(j) - Synchronization samples
772 j = 1;
773 for i=2:size(Sinc,1)-10
774     if((Sinc(i)-m(j))>10)
775         j = j+1;
776         m(j) = Sinc(i);
777     end
778 end
779 NN = length(m);                      % NN is the number of recorded
780 sweeps
781 clear Sinc i j
782
783 %% AER CALCULATION
784 for i=1:NN
785     AER = AER + Resp(m(i):m(i)+length(AER)-1)/NN; % Sweeps averaging
786 end
787 AER = AER(Phase_delay:length(AER)); % Phase distortion compensation
788 clear i
789
790 %% STORAGE
791 save(Name_File,'AER','EEG','m','NN','fs');
792 fprintf('Data in <%s.mat>\n',Name_File);
793
794

```

795

Table 1.

796

797

Element	Rough cost
Amplifier electronics <sup>1</sup>	200 USD
Electrodes and electrolytic paste	200 USD
Etymotic ER-3A insert earphones	500 USD
External AD/DA sound card	50 USD
TOTAL	950 USD

798

<sup>1</sup> Amplifier electronics include semiconductor elements, integrated circuits, connectors, PCB card, box, batteries, and battery holders.

799

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Table 2.

Evoked response	Bandwidth	High pass filter				Low pass filter			
		R1-H	R2-H	C1-H	C2-H	R1-L	R2-L	R1-H	R1-H
ABR	[150 3500]	33k $\Omega$	33k $\Omega$	47nF	22nF	6.8k $\Omega$	6.8k $\Omega$	4.7nF	10nF
MLR	[0.5 3500]	1M $\Omega$	1M $\Omega$	470nF	470nF	6.8k $\Omega$	6.8k $\Omega$	4.7nF	10nF

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Table 3.

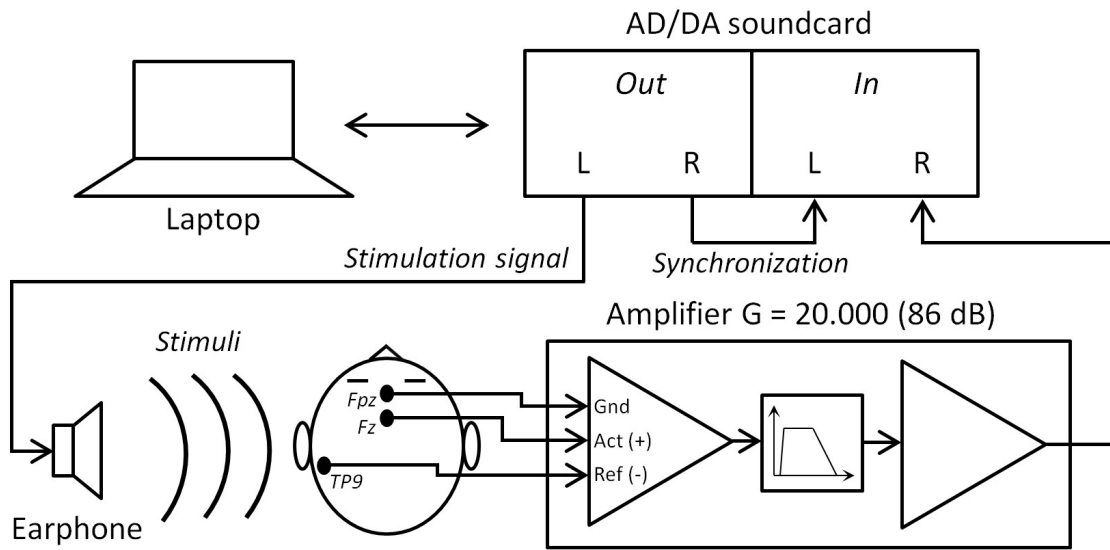
Feature	Value
Sampling rate	25 kHz
Input range	-3 V / +3 V
Output range	-2.5 V / +2.5 V
Bits of quantization	16
Quantization step	91.55 $\mu$ V

806

807

ACCEPTED MANUSCRIPT

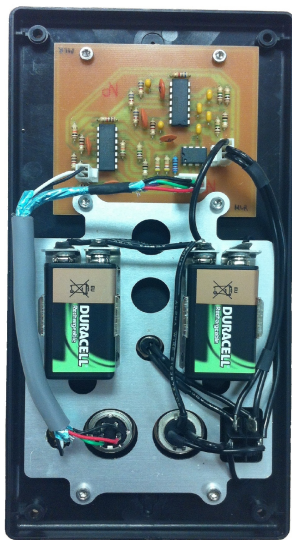
808 Figure 1



809

810

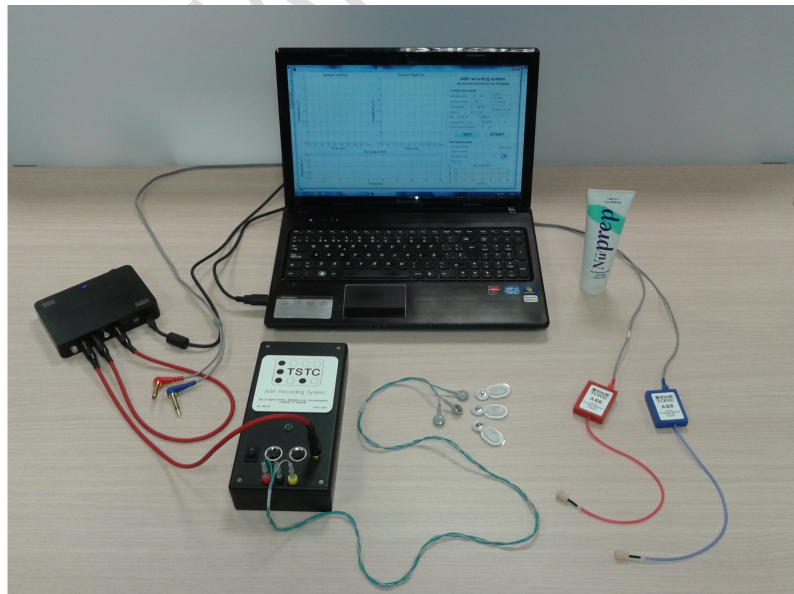
811 Figure 2



812

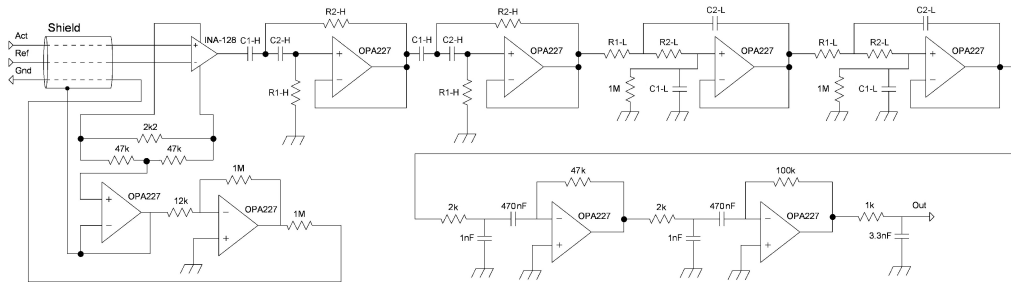
813

814





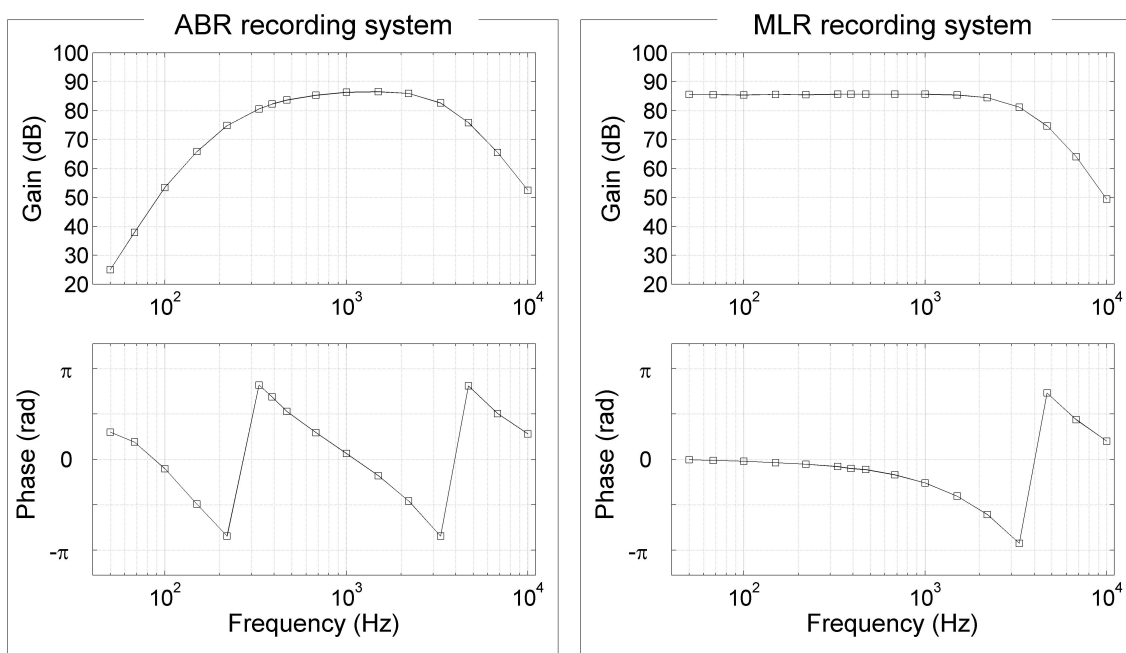
815 Figure 3



816

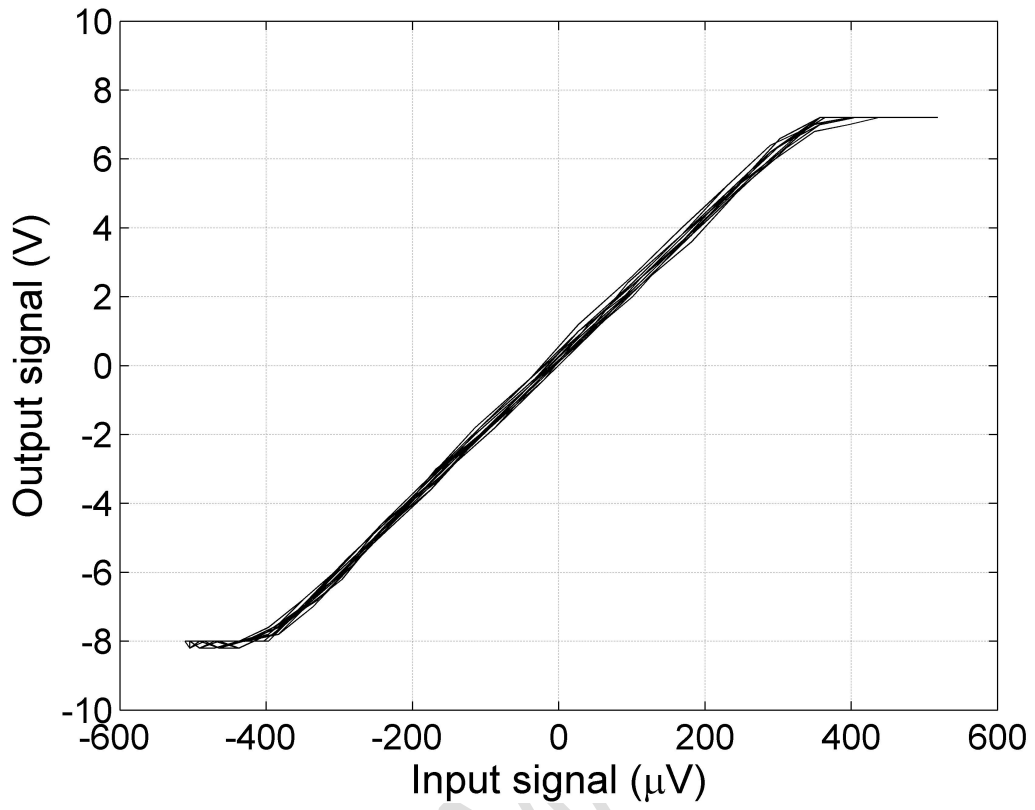
817

818 Figure 4



819

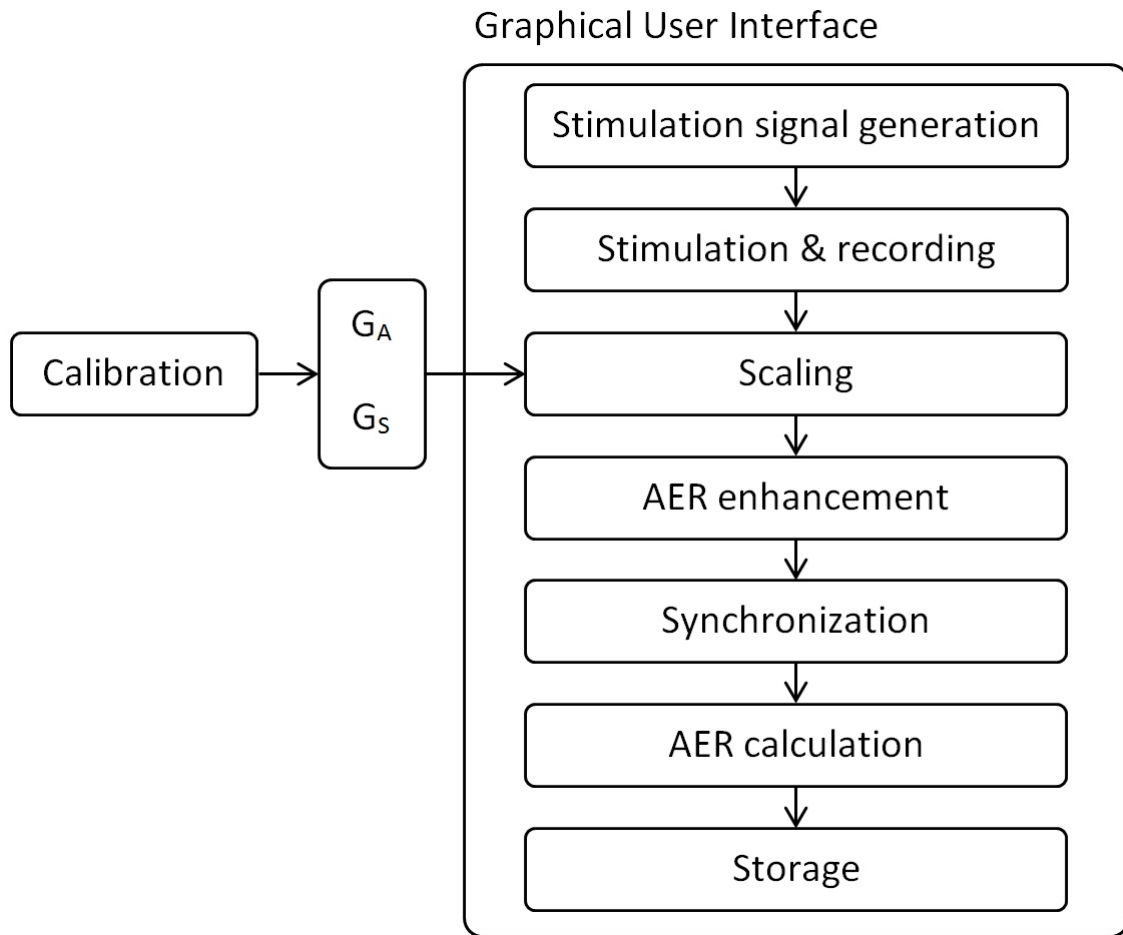
820 Figure 5



821

822

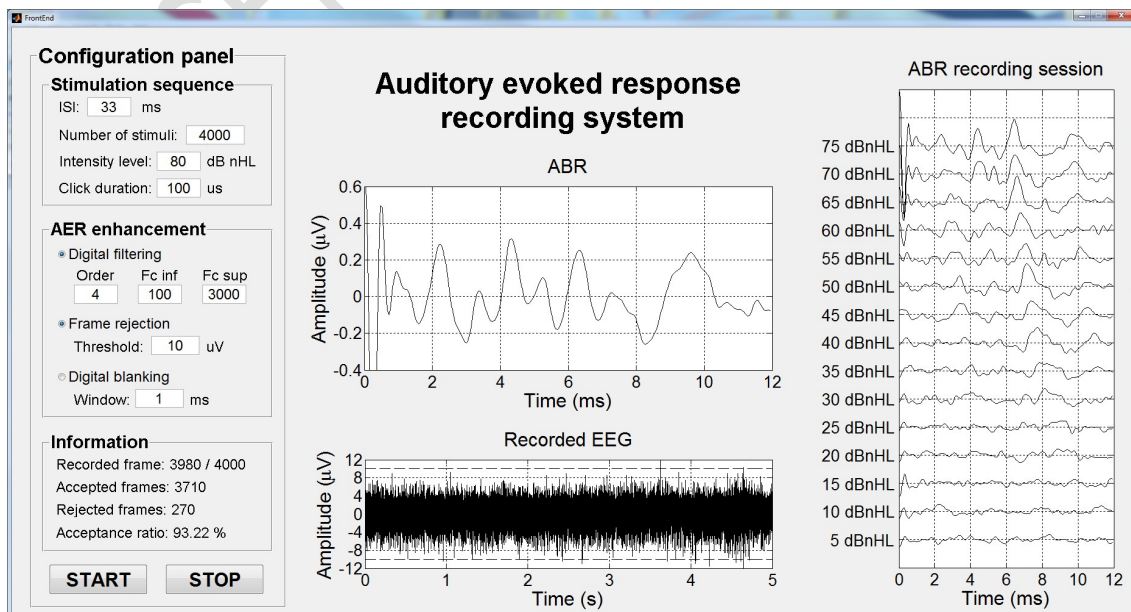
ACCEPTED MANUSCRIPT



824

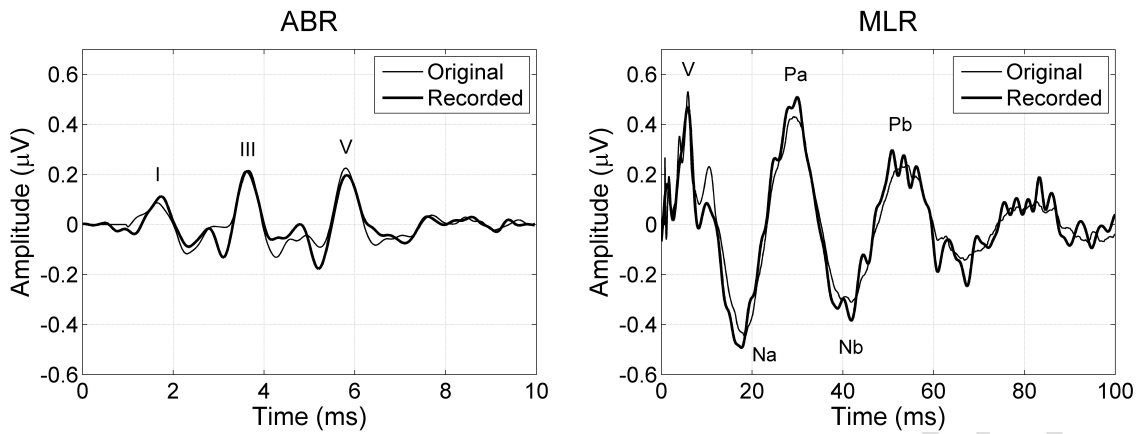
825

826 Figure 7



827

828 Figure 8

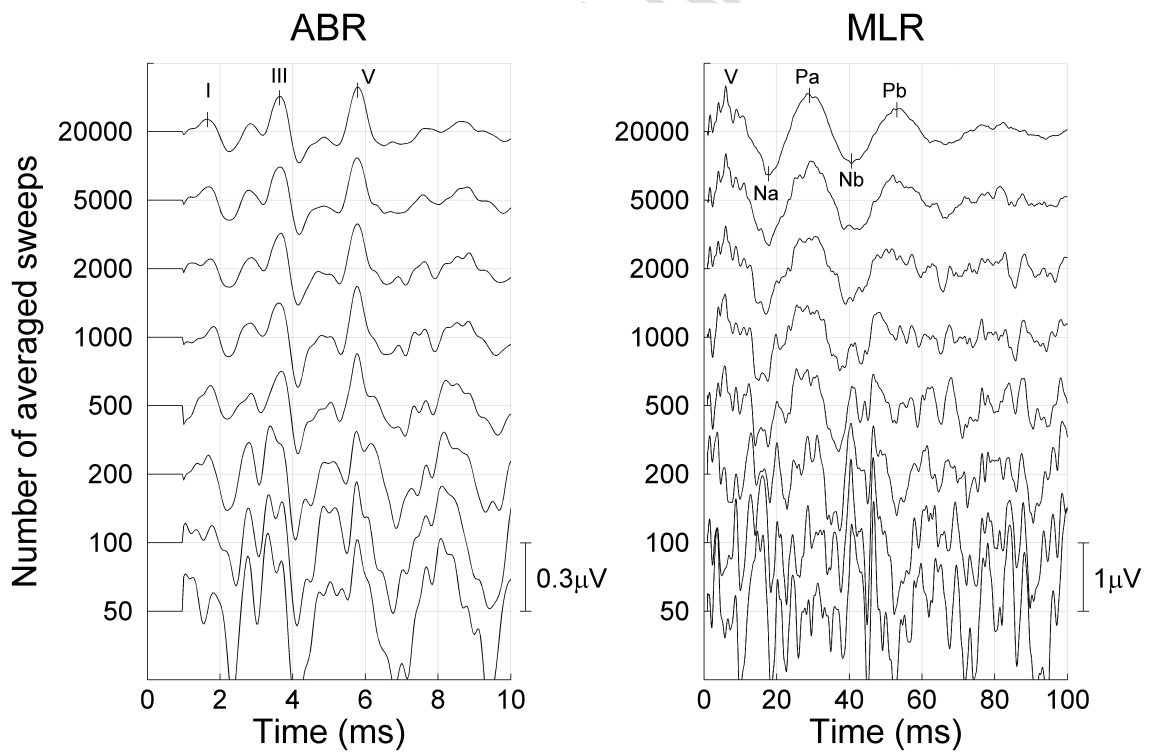


829

830

831 Figure 9

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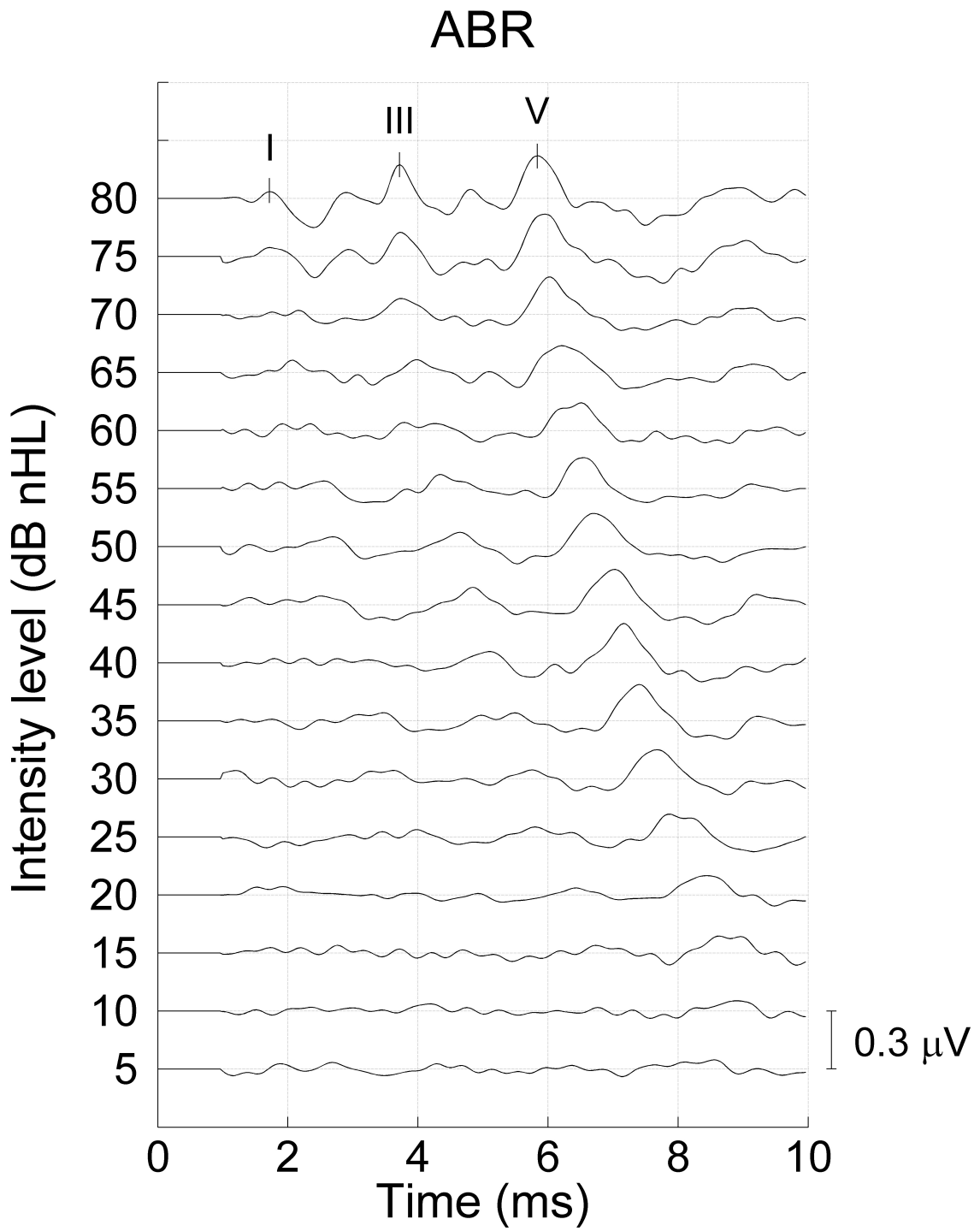


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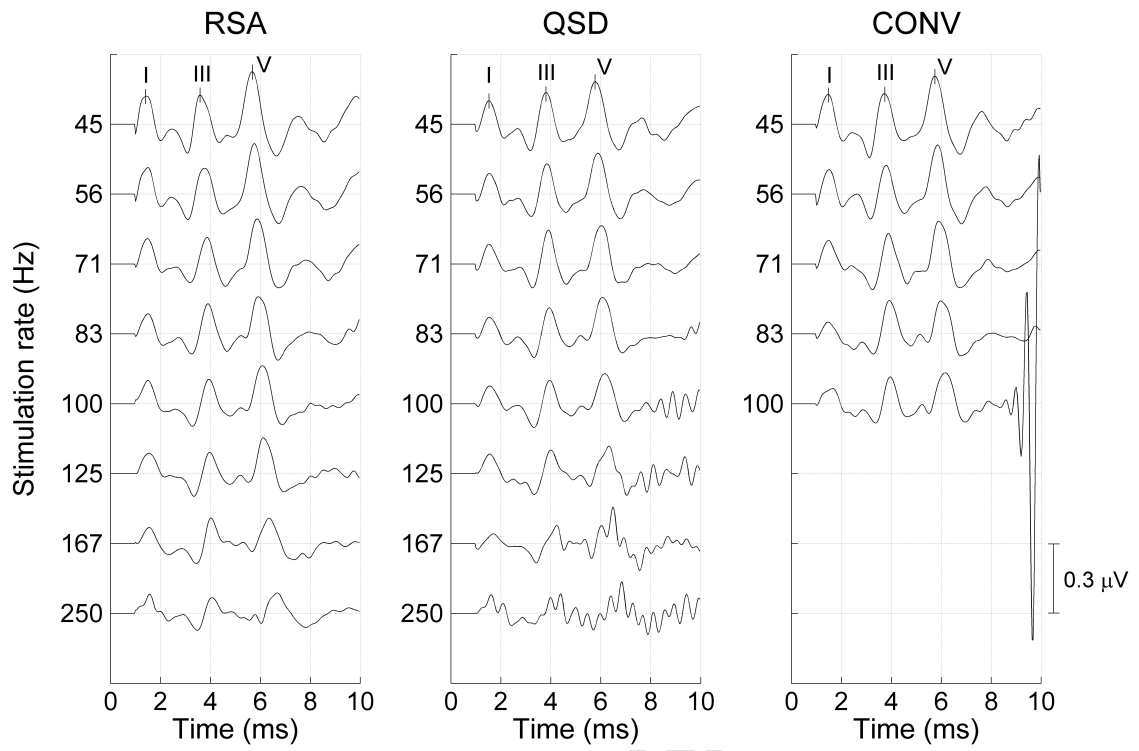


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841 Figure 11



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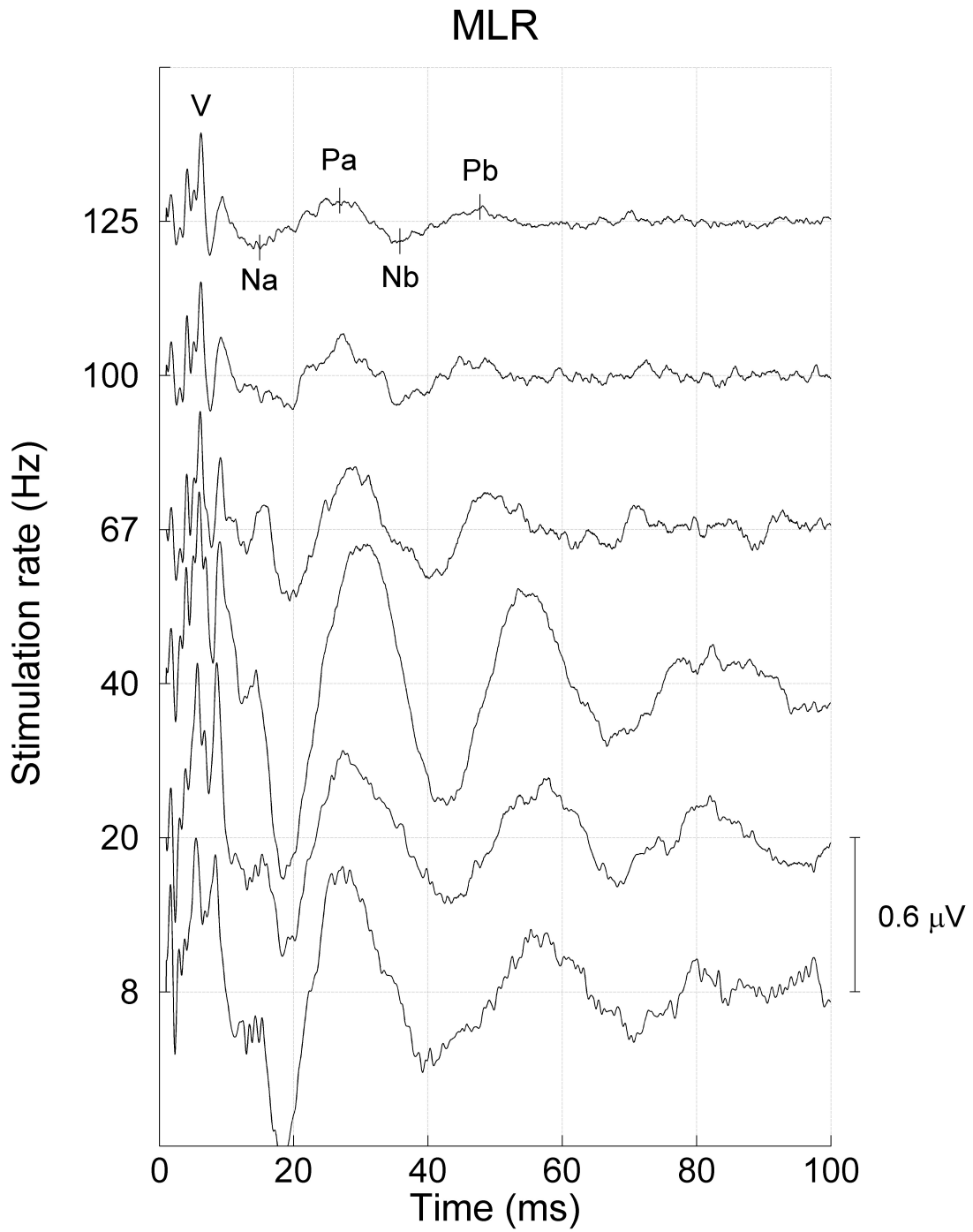
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ACCEPTED MANUSCRIPT

845 Figure 12

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