

Latency-dependent filtering and compact representation of the complete auditory pathway response

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1 Auditory evoked potentials (AEPs) include the auditory brainstem response (ABR),
2 middle latency response (MLR) and cortical auditory evoked potentials (CAEPs),
3 each one covering a specific latency range and frequency band. For this reason, ABR,
4 MLR and CAEP are usually recorded separately using different protocols. This arti-
5 cle proposes a procedure providing a latency-dependent filtering and down-sampling
6 of the AEP responses. This way, each AEP component is appropriately filtered, ac-
7 cording to its latency, and the complete auditory pathway response is conveniently
8 represented (with the minimum number of samples, i.e., without unnecessary redun-
9 dancies). The compact representation of the complete response facilitates a compre-
10 hensive analysis of the evoked potentials (keeping the natural continuity related to the
11 neural activity transmission along the auditory pathway), which provides a new per-
12 spective in the design and analysis of AEP experiments. Additionally, the proposed
13 compact representation reduces the storage or transmission requirements when large
14 databases are manipulated for clinical or research purposes. The analysis of the AEP
15 responses shows that a compact representation with 40 samples/decade (around 120
16 samples) is enough for accurately representing the response of the complete auditory
17 pathway and provides appropriate latency-dependent filtering. MatLab/Octave code
18 implementing the proposed procedure is included in the supplementary materials.

Keywords: Auditory Evoked Potentials (AEPs); auditory pathway; electroencephalo-
gram (EEG); auditory brainstem response (ABR); middle latency response (MLR);
cortical auditory evoked potentials (CAEPs).

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19 **I. INTRODUCTION**

20 Auditory evoked potentials (AEPs) are registered by presenting an auditory stimulus and
21 recording the neural activity elicited by the stimulus. Due to the noise affecting the recording
22 procedure and the low amplitude of the responses (typically in the range of microvolts), a
23 number of responses are synchronously averaged in order to improve the signal to noise ratio
24 (SNR) (Thornton, 2007).

25 Conventional recording procedures configure a minimal separation between consecutive
26 stimuli greater than the response duration in order to avoid interference among adjacent re-
27 sponses (Woldorff, 1993). This requirement has conditioned the protocols for recording AEP
28 responses, with different configurations for each portion of the response. For instance, the
29 auditory brainstem response (ABR) includes waves I, II, III, IV, V and VII, with latencies
30 in the range 1-10 ms. Conventional ABR recording protocols configure an inter-stimulus
31 interval (ISI) greater than 15 ms, and a band-pass filtering of the electroencephalogram
32 (EEG) in the band 100-3 000 Hz, removing the later responses (in the frequency band below
33 100 Hz) as well as the high frequency noise (above 3 kHz) (Burkard and Don, 2007). Simi-
34 larly, the middle latency response (MLR) includes waves N_0 , P_0 , N_a , P_a , N_b and P_b in the
35 latency range 10-100 ms. Therefore, recording windows of 100 ms, ISI greater than 120 ms
36 and EEG band-pass filtering in the band 10-300 Hz are considered in the conventional MLR
37 recording protocol (Pratt, 2007). Finally, the cortical auditory evoked potentials (CAEP),
38 with waves P_1 , N_1 , P_2 , N_2 and P_3 between 50 and 500 ms, are conventionally recorded with
39 protocols using recording windows of 1 s, ISI greater than this window and EEG band-pass

40 filtering in the band 1-30 Hz (removing high frequency noise as well as earlier responses)
41 (Martin *et al.*, 2007). The bandwidth limit (due to the band-pass filtering) allows EEG
42 acquisition at appropriate sampling rates (10 kHz, 1 kHz and 100 Hz for ABR, MLR and
43 CAEP, respectively) without information loss, according to the sampling theorem.

44 In the last decades, AEPs evoked by stimuli presented at high rate have offered new
45 perspectives in audiology. The possibility of recovering the evoked response when the in-
46 terval between stimuli is shorter than the response duration allows the study of different
47 adaptation mechanisms (Gillespie and Muller, 2009; Thornton and Coleman, 1975) as well
48 as the analysis of the AEP response to progressively more natural stimuli (Maddox and Lee,
49 2018). Some procedures have been proposed for recording evoked potentials at high stim-
50 ulation rates: maximum length sequences (MLS) (Eysholdt and Schreiner, 1982; Thornton
51 and Slaven, 1993), adjacent-responses (ADJAR) (Woldorff, 1993), quasi-periodic sequence
52 deconvolution (QSD) (Jewett *et al.*, 2004), continuous loop averaging deconvolution (CLAD)
53 (Bohorquez and Ozdamar, 2006; Ozdamar and Bohorquez, 2006), linear deconvolution for
54 baseline correction (LDBC) (Lütkenhöner, 2010), randomized stimulation and averaging
55 (RSA) (Valderrama *et al.*, 2012), iterative randomized stimulation and averaging (IRSA)
56 (de la Torre *et al.*, 2019; Valderrama *et al.*, 2014b, 2016) and least-squares deconvolution
57 (LS) (Bardy *et al.*, 2014a,b).

58 The overcoming of the ISI restriction allows the simultaneous recording of the different
59 portions of the AEP response (ABR, MLR and CAEP) at moderate or high stimulation rate
60 (de la Torre *et al.*, 2019; Holt and Ozdamar, 2016; Kohl *et al.*, 2019). The analysis of such
61 response provides information about the whole auditory pathway taking into consideration

62 all the involved waves, instead of a separated analysis of the different groups of waves in
63 different representations. The simultaneous analysis of all the waves (and their changes
64 associated to the modification of the stimulation parameters) provides new perspectives in
65 the study of the auditory system and its response to stimulation patterns progressively more
66 complex, from those so simple as quasi-periodic sequences of clicks to those so complex
67 as natural speech. Additionally, the analysis of the complete auditory pathway response
68 eliminates the discontinuity usually established among the different groups of waves. This
69 discontinuity does not exist in the generation of the neural activity and is a consequence of
70 the conventional protocols for the acquisition of the evoked responses.

71 The acquisition of evoked responses from the whole auditory pathway (including ABR,
72 MLR and CAEP) presents some difficulties. On one hand, the appropriate filtering is dif-
73 ferent for each portion. This is usually solved by applying the less restrictive filtering to the
74 EEG (using the band 1-3000 Hz), but the late portion of the response is more affected by
75 noise (compared with conventional recording procedures). On the other hand, the represen-
76 tation of the whole response requires a high sampling rate associated to the bandwidth of the
77 earlier waves (10 kHz) and a long duration of the response associated to the latencies of the
78 later waves (1 000 ms), which implies a large number of samples to represent the response
79 (typically around 10 000 samples) (de la Torre *et al.*, 2019). In order to illustrate the highly
80 redundant representation of the whole response we can compare these 10 000 samples with
81 those required for representing each portion separately: 100 samples for ABR (10 ms at a
82 sampling rate of 10 kHz), 100 samples for MLR (100 ms at 1 kHz) and 100 samples for
83 CAEP (1 000 ms at 100 Hz), i.e. a total of 300 samples.

84 Of course, the reason behind these differences is the application of specific filtering and
85 sampling rate to each component (ABR, MLR and CAEP) when they are independently
86 represented, according to the expected frequency content (which changes with the latency).
87 This allows an specific filtering of each portion, providing appropriate noise reduction and
88 compact representation (independent for ABR, MLR and CAEP) but generates the discon-
89 tinuity in the representation of the auditory pathway response.

90 In general, a band-limited signal can be low-pass filtered (to remove high-frequency noise)
91 and sampled with a sampling rate at least twice the maximum frequency component (to re-
92 duce the number of samples required to represent it) without information loss, since the
93 sampling theorem guarantees that the original signal can be recovered from the samples.
94 Similarly, the evoked response of the complete auditory pathway, with a band limit depend-
95 ing on the latency (the later the waves the narrower bandwidth), can be processed to apply
96 a latency-dependent low-pass filtering and down-sampling, in order to improve both the
97 filtering (to reduce the high-frequency noise) and the representation (to reduce the number
98 of samples required for properly representing the response). This article proposes a digital
99 procedure to provide this latency-dependent low-pass filtering and down-sampling for the
100 evoked response of the complete auditory pathway. The procedure is based on a non-uniform
101 sampling (involving a compression of the latency axis) of the signal representing the AEP
102 response. The compression, approximately logarithmic, provides a bandwidth limitation in
103 terms of the maximum number of oscillations per decade, as well as the sampling rate in
104 terms of number of samples per decade (where a decade in the latency axis is the interval
105 between a latency t_0 and a latency $10 \cdot t_0$).

106 The proposed procedure is described with a matrix formulation, where the evoked re-
 107 sponse is represented as a signal with J samples (or a J -component column vector), the
 108 reduced representation of the evoked response (after the latency-dependent filtering and
 109 down-sampling) as a signal with J_r samples (or a J_r -component column vector, with $J_r < J$),
 110 and the latency-dependent filtering and down-sampling procedure is represented as a matrix
 111 with J_r rows and J columns. The matrix processes the original AEP response and provides
 112 its reduced representation, and also allows to recover, from the reduced representation, the
 113 filtered AEP response in the original representation (i.e., from the compact representation
 114 with J_r samples, it provides the signal in the conventional representation with J samples at
 115 the original sampling rate, including the latency-dependent filtering).

116 The proposed procedure is described in this article and MatLab/Octave code to gener-
 117 ate the latency-dependent low-pass filtering and down-sampling matrix is provided in the
 118 supplementary materials. The procedure has been evaluated in experiments involving the si-
 119 multaneous recording of ABR, MLR and CAEP using clicks as stimuli, presented at different
 120 stimulation rates.

121 **II. FORMULATION OF THE LATENCY-DEPENDENT LOW-PASS FILTERING** 122 **AND DOWN-SAMPLING**

123 **A. Matrix formulation of filtering and down-sampling**

124 Let's suppose a digital signal $x(j)$ representing an AEP response. Low-pass filtering is
 125 obtained as the convolution of the signal $x(j)$ with the impulsive response of the filter, $h_{lp}(j)$,

126 as:

$$x_{lp}(j) = h_{lp}(j) * x(j) = \sum_{j'} h_{lp}(j') \cdot x(j - j') \quad (1)$$

127 where ‘*’ represents convolution. If the signal $x(j)$ contains J samples ($j = 0, \dots, J - 1$), it
 128 can be represented as a J -component column vector, and filtering can be represented as a
 129 matrix product:

$$\mathbf{x}_{lp} = H_{lp}\mathbf{x} \quad (2)$$

130 where \mathbf{x}_{lp} is a J -component column vector representing the filtered signal and H_{lp} is the
 131 $J \times J$ convolution matrix with elements $H_{lp}(j_1, j_2) = h_{lp}(j_1 - j_2)$. The matrix product can
 132 be interpreted as a linear operator in the J -dimensional vectorial space representing the
 133 digital signals: each filtered sample $x_{lp}(j)$ is obtained as a linear combination of the samples
 134 of the original signal, according to the j^{th} row of the convolution matrix.

135 The low-pass filtered signal can be down-sampled without information loss if the new
 136 sampling rate is, at least, twice the maximum frequency component (according to the sam-
 137 pling theorem). In order to down-sample the filtered signal with a factor q , from every q
 138 samples 1 sample should be preserved and $q - 1$ samples should be discarded:

$$x_r(j_r) = x_{lp}(j_r \cdot q) \quad (3)$$

139 where $j_r = 0, \dots, J_r - 1$ (with $J_r = J/q$), and r refers to the reduced representation after
 140 down-sampling. The low-pass filtering and down-sampling operation can also be represented
 141 in matrix notation:

$$\mathbf{x}_r = H_r\mathbf{x} \quad (4)$$

142 where H_r is the matrix H_{lp} preserving 1 of every q rows (i.e., a $J_r \times J$ matrix). Again, this
 143 matrix product can be considered a linear transformation from the original representation
 144 of the signal (with J components) to a reduced representation (with J_r components), where
 145 each new component $x_r(j_r)$ is obtained as a linear combination of all the original components
 146 $x(j)$ according to the j_r^{th} row of the matrix H_r . The supplementary materials¹ (Section 1)
 147 describe in detail the matrix formulation of filtering and down-sampling.

148 **B. Latency-dependent filtering and down-sampling**

149 A latency-dependent filtering implies that each filtered component $x_{lp}(j)$, is obtained
 150 using a different impulsive response, depending on its latency. In other words, while in
 151 conventional filtering all the rows of the matrix H_{lp} are identical except for the delay (and all
 152 the elements in each direct diagonal of the matrix are identical), latency-dependent filtering
 153 can be implemented using a different impulsive response for each row of the matrix. This
 154 way, each filtered sample at latency j can be obtained with a latency-specific bandwidth
 155 (using an appropriate impulsive response) as a linear combination of the original samples
 156 around this latency.

157 Similarly, latency specific down-sampling can be implemented by selecting the new sam-
 158 ples with a latency-dependent down-sampling factor q (according to the latency-dependent
 159 bandwidth, in order to locally follow the sampling theorem condition and prevent infor-
 160 mation loss). The latency-dependent down-sampling can also be implemented as a matrix
 161 operation, using a reduced matrix H_r in which the rows are non-uniformly selected from

162 the matrix H_{lp} . Section 2 of the supplementary materials¹ describes in detail the matrix
163 formulation of the latency-dependent filtering and down-sampling.

164 C. Bandwidth required at each latency

165 The latency-dependent low-pass filtering and down-sampling has to be designed taking
166 into account the expected frequency content (and therefore the required bandwidth) of the
167 AEP responses at each latency. The different waves of the AEP responses are narrower
168 at earlier latency and wider at later latency, and the required bandwidth decreases as the
169 latency increases. Taking into account the latency range of the waves and the typical cut-off
170 frequencies used for recording ABR, MLR and CAEP responses, the required bandwidth can
171 be determined for each latency. The waves of ABR are observed at latencies between 1 and
172 10 ms, and recording protocols apply typically a 3 kHz low-pass filtering in order to preserve
173 the waves and reduce the high frequency noise. Similarly, MLR recording procedures, with
174 waves between 10 ms and 100 ms typically apply 300 Hz low-pass filtering. Finally, CAEP
175 are usually recorded with a 30 Hz low-pass filtering to preserve waves in the latencies between
176 50 ms and 1 s and remove the high frequency noise. Therefore, a latency-dependent filtering
177 preserving a bandwidth of 3 KHz at 1 ms, 300 Hz at 10 ms, and 30 Hz at 50 ms would be
178 enough for an appropriate representation of the AEP responses.

179 D. Compression of the latency axis

180 The latency-dependent low-pass filtering and down-sampling can be implemented as a
181 uniform low-pass filtering and down-sampling performed after a compression of the latency

182 axis. The progressive reduction of the bandwidth with the latency suggests a logarithmic
 183 scaling of the latency axis, which would be described as a non-uniform sampling with K_{dec}
 184 samples per decade (i.e., a constant number of samples between a given latency t_0 and a
 185 latency $10 \cdot t_0$). The response in the original representation contains samples at the time
 186 values:

$$t_j = j T_s \quad (5)$$

187 where $T_s = 1/f_s$ is the sampling period and f_s is the sampling rate. The uniform sampling in
 188 a logarithmically compressed latency axis, for K_{dec} samples per decade, would be described
 189 with the equation:

$$j_r(t) = K_{dec} \log_{10}(t/T_s) \quad (6)$$

190 where the samples should be taken at those values of t providing an integer value of $j_r(t)$.
 191 This latency compression would be appropriate for large latencies (much greater than T_s)
 192 (since an increment of t in a factor 10 would produce an increment of K_{dec} samples, providing
 193 K_{dec} samples per decade). However, it is not appropriate at small latency, because the
 194 sampling rate would be very large when t is small compared with T_s . Instead, a linear-
 195 logarithmic (lin-log) compression can be applied to the latency axis, providing a linear
 196 sampling at small latency (compared with the original sampling period) and a logarithmic
 197 sampling at large latency. The equation providing the relation between the original time axis
 198 and the compressed samples, with a linear compression at small latency (with a maximum
 199 sampling rate equal to f_s) and K_{dec} samples per decade at large latency, is:

$$j_r(t) = K_{dec} \log_{10} \left(\frac{t}{T_s} \frac{\ln(10)}{K_{dec}} + 1 \right) \quad (7)$$

latency	$K_{dec} = 40$ samp/dec		$K_{dec} = 60$ samp/dec	
	$f_s=14.7$ kHz	$f_s=25$ kHz	$f_s=14.7$ kHz	$f_s=25$ kHz
1 ms	7.96 kHz	10.25 kHz	9.40 kHz	12.76 kHz
2 ms	5.46 kHz	6.45 kHz	6.91 kHz	8.57 kHz
5 ms	2.81 kHz	3.05 kHz	3.85 kHz	4.31 kHz
10 ms	1.55 kHz	1.62 kHz	2.21 kHz	2.36 kHz
20 ms	820.1 Hz	839.4 Hz	1.20 kHz	1.24 kHz
50 ms	339.4 Hz	342.7 Hz	503.3 Hz	510.5 Hz
100 ms	171.7 Hz	172.5 Hz	256.0 Hz	257.9 Hz
200 ms	86.3 Hz	86.6 Hz	129.1 Hz	129.6 Hz
500 ms	34.7 Hz	34.7 Hz	51.9 Hz	52.0 Hz
1000 ms	17.4 Hz	17.4 Hz	26.0 Hz	26.0 Hz

TABLE I. Latency-dependent sampling rate $f'_s(t)$ for different original sampling rates (f_s) and resolutions (K_{dec}) in the compressed latency axis.

200 where $\ln()$ is the natural logarithm (the mathematical derivation of this equation is included
201 in the supplementary materials¹, Section 3). The sampling period can be estimated from
202 this equation as the derivative $\partial t(j_r)/\partial j_r$, or equivalently as the inverse of the derivative
203 $\partial j_r(t)/\partial t$. The latency-dependent sampling period and sampling rate are, respectively:

$$T'_s(t) = \left. \frac{\partial t(j_r)}{\partial j_r} \right|_t = T_s + t \frac{\ln(10)}{K_{dec}} \quad f'_s(t) = \frac{1}{T'_s(t)} \quad (8)$$

204 As can be observed from this equation, for small latency ($t \ll T_s$), the sampling period is
205 minimum and equal to T_s (the sampling rate takes the maximum value, f_s) and for large
206 latency ($t \gg T_s$), the sampling rate is $f'_s(t) \approx K_{dec}/(t \cdot \ln(10))$, and therefore it decreases as
207 the latency increases, and depends on the latency but not on the original sampling rate.

208 The bandwidth preserved at each latency depends on the local sampling rate and the
209 frequency response of the latency-dependent low-pass filter. Even though the sampling

210 theorem limits the bandwidth to half of the sampling rate, i.e., $f'_s(t)/2$, a slightly smaller
 211 bandwidth is recommended in order to allow a reasonable implementation of the filters
 212 (otherwise the duration of the impulsive response would be too long). Table I shows the
 213 local sampling rate $f'_s(t)$ for original sampling rates of 14.7 and 25 kHz, and for resolutions
 214 of 40 and 60 samples/decade. All these configurations provide enough bandwidth for the
 215 representation of the AEP responses.

216 E. Design of the low-pass filters

217 For the latency-dependent low-pass filtering, a root raised-cosine (RRC) filter in the com-
 218 pressed latency axis has been designed (Proakis and Salehi, 2008). Filters in the RRC family
 219 (commonly used in digital communications) are low pass filters specified by two parameters:
 220 the symbol period T_0 and the roll-off factor α . They provide a constant frequency response
 221 up to $(1 - \alpha)/(2T_0)$, a monotonic decay up to $(1 + \alpha)/(2T_0)$ and a null response for frequen-
 222 cies above this value. Although its theoretical impulsive response is infinite, a truncated
 223 version of an RRC filter can be implemented as a linear phase FIR (Finite Impulsive Re-
 224 sponse) using a time span including a sufficient number of symbol periods. Additionally,
 225 RRC responses are orthogonal when they are delayed an integer number of symbol peri-
 226 ods. Detailed information about RRC filters is provided in the supplementary materials¹
 227 (Section 4). In this work, RRC filters with a roll-off $\alpha=0.2$ are used with an impulsive
 228 response length of ± 14 symbol periods, containing the 99.9980% of the total energy of the
 229 theoretical impulsive response. Its frequency response is constant in the range $[0, 0.4/T_0]$,
 230 monotonically decreasing in the range $[0.4/T_0, 0.6/T_0]$ and null for frequency greater than

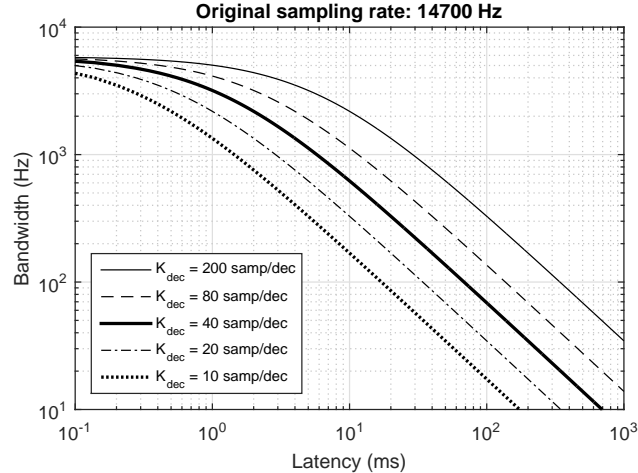


FIG. 1. Bandwidth preserved by the latency dependent filtering using RRC filters with $\alpha = 0.2$.

231 $0.6/T_0$. A non-causal zero-phase implementation of the filters has been considered in order
 232 to avoid delay of the filtered AEP waves.

233 In order to achieve the latency-dependent low-pass filtering, the time axis of the im-
 234 pulsive response is scaled according to equation (7) (the impulsive response is invariant in
 235 the compressed latency axis given by j_r). This latency-dependent low-pass filtering can be
 236 represented as a matrix operation H_{lp} where the impulsive response for each latency (repre-
 237 sented by each row) is wider as the latency increases (i.e., as we move from top to bottom in
 238 the matrix rows). Similarly, the latency-dependent down-sampling can be represented as a
 239 reduced matrix H_r , obtained from H_{lp} where only those rows corresponding to the latencies
 240 $t(j_r)$ (with j_r integer) are selected.

241 The RRC filters are designed with a symbol period T_0 matching the sampling period $T'_s(t)$,
 242 constant in the compressed latency axis and therefore increasing with the latency according
 243 to equation (8). This preserves (without distortion) the frequency range $[0, 0.4] \cdot f'_s$, removes
 244 all the frequency components above $0.6 \cdot f'_s$, and produces interference (by aliasing) with the

245 frequency components in the range $[0.4, 0.6] \cdot f'_s$. If the signal of interest contains, at latency
 246 t , only frequency components below $0.4 \cdot f'_s(t)$, the signal is preserved without distortion, and
 247 the aliasing only affects the noise components in the range $[0.4, 0.6] \cdot f'_s$. Figure 1 represents
 248 the bandwidth preserved by the latency dependent filtering using the proposed RRC filters,
 249 for different resolutions (K_{dec} between 10 and 200 samples/decade) and original sampling
 250 rate $f_s=14\,700$ Hz. For instance, for $K_{dec}=40$ samples/decade, the preserved bandwidth is
 251 3 185, 621, 68.6 and 6.94 Hz at latencies 1, 10, 100 and 1000 ms, respectively. Detailed
 252 description of the latency-dependent low-pass filtering and down-sampling using RRC filters
 253 is provided in the Section 5 of the supplementary materials¹.

254 **F. Orthonormalization of the latency-dependent filtering and down-sampling ma-** 255 **trix**

256 The use of a symbol period T_0 equal to the sampling period $T'_s(t)$ in the definition of
 257 the RRC filters provides orthogonality between the impulsive responses associated to each
 258 sample in the compressed latency axis. However, due to the non-linear compression of the
 259 latency axis, the impulsive responses are quasi-orthogonal but not orthogonal in the not-
 260 compressed latency axis. Orthonormalization (i.e., orthogonalization and normalization)
 261 of the matrix H_r providing the latency-dependent low-pass filtering and down-sampling is
 262 highly recommendable because, this way, the matrix provides an equivalent representation
 263 of the signals in the subspace of the band-limited signals that preserves the metric in the
 264 reduced representation subspace (i.e. the energies or the distances between signals in the
 265 original and the reduced representation space are invariant). The representation obtained
 266 with an orthonormal matrix is equivalent to the original representation, and therefore all the

267 estimations or algorithms can be equivalently performed either in the original J -dimensional
 268 representation space or in the J_r -dimensional reduced representation space (if the matrix was
 269 not orthonormal, the metric in the reduced representation space would be distorted and the
 270 results could not be equivalently obtained in the original and the reduced representations).

271 In order to orthonormalize the matrix H_r , a Gram-Schmidt process is applied. Since
 272 the number of remaining rows in the filtering and down-sampling matrix is significantly
 273 smaller than the number of columns, an orthonormalization based on Gaussian elimination is
 274 proposed. After the orthonormalization, the rows of the matrix V_r constitute an orthonormal
 275 basis of functions describing the subspace of the latency-dependent band-limited signals. If
 276 V_r is the matrix resulting from the orthonormalization of the H_r matrix, then the product
 277 $V_r V_r^T$ (where V_r^T is the transposed of V_r) is the $J_r \times J_r$ identity matrix. The orthonormal
 278 matrix V_r can be used to project the original signal to the subspace of the latency-dependent
 279 band-limited signals:

$$\mathbf{x}_r = V_r \mathbf{x} \quad (9)$$

280 This matrix operation removes all the components out of the subspace defined by the basis
 281 of functions and provides a compact representation of those components within the subspace
 282 of the latency-dependent band limited signals. A MatLab/Octave function providing the
 283 orthonormalized latency-dependent low-pass filtering and down-sampling matrix V_r has been
 284 implemented (included in the supplementary materials¹, Section 6). This function includes
 285 the compression of the latency axis according to equation (7), the definition of J_r responses
 286 uniformly distributed in the compressed latency axis and the orthonormalization of the

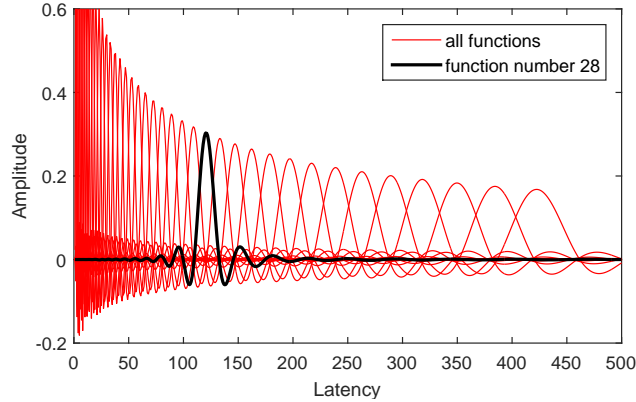


FIG. 2. (Color online) Functions of the orthonormalized basis for $J=500$ samples and $K_{dec}=25$ samples/decade.

287 resulting functions in order to provide the matrix V_r containing the orthonormal basis of the
 288 reduced representation space.

289 Figure 2 represents the functions of the orthonormalized basis (i.e., the rows of the matrix
 290 V_r) for the latency-dependent low-pass filtering and down-sampling procedure designed for
 291 $J=500$ samples and $K_{dec}=25$ samples/decade (resulting in $J_r=41$). The RRC shape (in the
 292 compressed latency axis) can be appreciated in the function represented with the thicker line
 293 (corresponding to the 28-th function). Section 7 of the supplementary materials¹ includes
 294 some examples of the sampling functions in the matrices H_r and V_r (i.e., before and after
 295 the orthonormalization).

296 G. Reconstruction of the signal in the original representation

297 Since the matrix V_r is orthonormal, it can be directly applied to transform the reduced
 298 representation to the original representation:

$$\mathbf{x}_{lp} = V_r^T \cdot \mathbf{x}_r = V_r^T \cdot V_r \cdot \mathbf{x} \quad (10)$$

299 This matrix operation provides the latency-dependent filtered signal in the original repre-
 300 sentation (i.e., at the original sampling rate). Section 8 of the supplementary materials¹
 301 includes some examples comparing the \mathbf{x}_{lp} signal recovered from the previous equation and
 302 that obtained by filtering with H_{lp} . Slight differences associated to aliasing are also dis-
 303 cussed.

304 The representation of the latency-dependent low-pass filtered response at the original
 305 sampling rate is highly redundant, because at late latency the bandwidth of the response
 306 is very small compared with the sampling rate. Instead, the response can be reconstructed
 307 at specific latencies. In order to estimate the filtered signal at a specific latency t_i , the
 308 contribution of each component in the reduced representation must be considered:

$$x_{lp}(t_i) = \sum_{j_r=0}^{J_r-1} x_r(j_r) V_r(j_r, j(t_i)) \quad (11)$$

309 where $x_r(j_r)$ is the j_r -th component of the reduced representation \mathbf{x}_r and $V_r(j_r, j(t_i))$ is
 310 the j_r -th function of the basis (given by the j_r -th row of the matrix V_r) evaluated (or
 311 interpolated) at the latency t_i . This way, from the reduced representation, the filtered
 312 response $x_{lp}(t)$ can be evaluated at a reasonable set of latencies $\{t_i\}$ (for example, with
 313 200 samples per decade in the interval between 1 ms to 1 s) providing a representation
 314 more natural than the reduced representation \mathbf{x}_r (which modifies the amplitudes depending
 315 on the latency due to the orthonormalization of the matrix) and more compact than the
 316 representation at the original sampling rate.

317 The reconstruction of the response at specific latencies has been implemented as a ma-
 318 trix operation. A MatLab/Octave function providing the reconstruction matrix for specific

latencies is included in the supplementary materials¹ (Section 9). Some examples are also included.

III. EXPERIMENTAL RESULTS

The proposed latency-dependent filtering and down-sampling has two objectives: on one hand, to provide a filtering adapted to the AEP spectral content (which changes with the latency) in order to appropriately reduce the high frequency noise; on the other hand, to provide a compact representation of the AEP responses, with a sampling rate adapted to the spectral content (and therefore also changing with the latency), in order to reduce the number of samples required for representing the responses.

The evaluation of the proposed procedure has been performed using both simulations and real AEP responses. Simulations are based on synthetic EEGs, generated with an AEP response used as reference and contaminated with noise. Therefore, since the reference response is available, the noise affecting the estimated responses (either the not-filtered or the latency-dependent filtered) can accurately be evaluated in terms of the SNR. An evaluation of the noise reduction provided by the latency-dependent filtering is more difficult in experiments with recorded EEGs (since the reference response is not available) but provides more realistic results.

A. Experimental design

For the experiments involving simulations, an AEP response has been prepared as reference. This response corresponds to the grand average (estimated from 4 subjects) of an

339 AEP response to 0.1 ms rarefaction clicks presented at 74 dB (hearing level) at an average
 340 rate of 1.39 Hz, with a random inter-stimulus interval (ISI) with uniform distribution in the
 341 range 480-960 ms. The acquisition procedure and the response is described in detail in (de la
 342 Torre *et al.*, 2019). In spite of the grand average process this response contains some noise
 343 (particularly at late latency) due to the noise contaminating the EEGs. For this reason, the
 344 AEP response was latency-dependent filtered with a resolution of 40 samples/decade, using
 345 the corresponding orthonormalized matrix V_r :

$$\mathbf{x}_{ref} = V_r^T \cdot (V_r \cdot \mathbf{x}_0) \quad (12)$$

346 where \mathbf{x}_0 is the original grand-average AEP response and \mathbf{x}_{ref} is the filtered response used
 347 as reference. This resolution has been selected taking into account the spectral content of
 348 the AEP responses expected at each latency. The AEP response \mathbf{x}_{ref} (with 14 700 samples
 349 at a sampling rate 14.7 kHz, i.e., corresponding to a response length of 1 s.), is described in
 350 detail in the supplementary materials¹ (Section 10).

351 The reference response \mathbf{x}_{ref} was used to generate a synthetic EEG using a random ISI
 352 with uniform distribution in the interval 480-960 ms. The EEG was contaminated with
 353 pink noise (i.e., with power spectral density decreasing with 3 dB/octave), with a level
 354 providing a AEP response with a SNR around 10 dB (which is a typical noise level in
 355 AEP experiments). The AEP response \mathbf{x} was estimated from the synthetic EEG using the
 356 IRSA algorithm (de la Torre *et al.*, 2019; Valderrama *et al.*, 2014b, 2016). The latency-
 357 dependent low-pass filtering was applied to the response with resolutions between 10 and
 358 200 samples/decade ($\mathbf{x}_{lp} = V_r^T \cdot V_r \cdot \mathbf{x}$). The SNR was evaluated as the ratio between
 359 the energy of the reference response \mathbf{x}_{ref} and the energy of the noise \mathbf{n} contaminating the

360 evaluated response, i.e., $\mathbf{n} = \mathbf{x} - \mathbf{x}_{ref}$ for the non filtered response, $\mathbf{n} = \mathbf{x}_{lp} - \mathbf{x}_{ref}$ for the
 361 filtered responses. The required number of samples in the reduced representation $\mathbf{x}_r = V_r \cdot \mathbf{x}$
 362 was also evaluated.

363 In the experiments based on real EEGs, six stimulation rates ranging between 1.39 Hz
 364 (for ISI 480-960 ms) and 44.44 Hz (for ISI 15-30 ms) were considered. For each ISI condi-
 365 tion, 3 portions of EEG with 228 s were recorded (which accumulates 684 s). Rarefaction
 366 clicks of 0.1 ms presented at 74 dB (hearing level) were used as stimulation. The clicks were
 367 delivered diotically through ER-3A insert earphones. These transducers (using a delivery
 368 tube to separate the electromagnetic interference from the response) cause a group delay
 369 of around 1 ms (Elberling *et al.*, 2012). The response estimation is synchronized with the
 370 start of each stimulus in the transducer (which allows to appreciate the stimulation arti-
 371 fact at the beginning of the estimated responses), and therefore, since the group delay was
 372 not compensated, a delay of about 1 ms is expected in the waves of the evoked responses.
 373 The electrical response was recorded with surface electrodes located at Fz (upper forehead,
 374 active), Tp10 (right mastoid, reference) and Fpz (middle forehead, ground) using an in-
 375 strumentation preamplifier (gain 70 dB; bandwidth 1-3 500 Hz) (Valderrama *et al.*, 2013,
 376 2014a,b). The preamplified EEG was digitized (44 100 Hz sampling rate, 16 bits/sample),
 377 low-pass filtered (4 000 Hz cut-off frequency) and down-sampled in a factor 3 (14 700 Hz final
 378 sampling rate). Eye-blinking artifacts were eliminated with the iterative template matching
 379 and suppression algorithm (ITMS) (Valderrama *et al.*, 2018). Eight subjects (aged 26-58
 380 years) participated in this study. The protocol followed in this study is in accordance with
 381 the Code of Ethics of the World Medical Association (Declaration of Helsinki) for exper-

382 iments involving humans and it was approved by the Research Ethics Committee of the
383 University of Granada, reference 961/CEIH/2019. The EEG recordings used in this study
384 are an extension of the database used in a recent study (de la Torre *et al.*, 2019). More
385 details about the experimental procedure can be found in this reference.

386 As in the case of synthetic EEGs, the AEP responses were obtained from the real EEGs
387 with the IRSA algorithm (de la Torre *et al.*, 2019; Valderrama *et al.*, 2014b, 2016). The
388 latency-dependent low-pass filtering has been applied with resolutions between 5 and 200
389 samples/decade. In order to evaluate the quality improvement provided by the latency-
390 dependent filtering we have used the three AEP responses estimated from each EEG portion
391 of 228 s: for each subject and ISI condition, the average from the three responses, filtered
392 with $K_{dec}=40$ samples/decade, was used as reference and the SNR was estimated for each
393 individual response (using the corresponding reference). The resulting individual SNR esti-
394 mations have been averaged (across subjects, ISI conditions and repetitions). In addition to
395 this estimation of the SNR (independent for each subject), a grand-average-based SNR was
396 estimated: For each ISI condition, the grand-average across subjects from each of the EEG
397 portions of 228 s were used as individual responses, and the average of them, filtered with
398 $K_{dec}=40$ samples/decade, was used as reference. The SNR was estimated from each grand-
399 average response using the corresponding reference, and the resulting SNR estimations were
400 averaged across ISI conditions and repetitions. The utility of the SNR estimated with these
401 procedures is limited, but they provide an objective comparison of the effect of the latency
402 dependent filtering under two noise conditions (subject-based responses are more affected
403 by noise than grand-average-based responses).

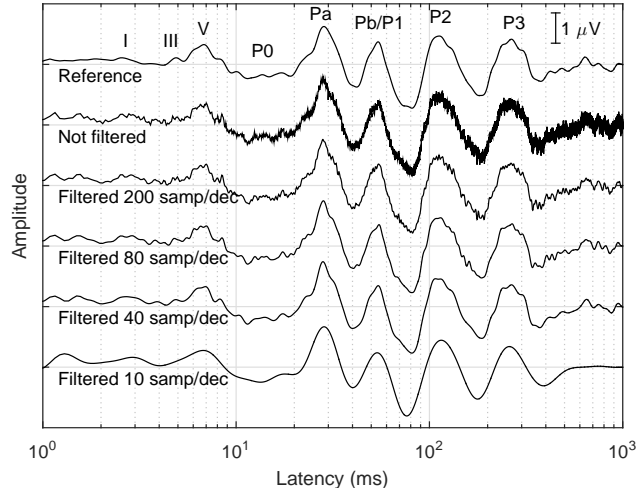


FIG. 3. From top to bottom, the reference AEP response, the IRSA estimation (without filtering) and the latency-dependent filtered responses with 200, 80, 40 and 10 samples/decade. Simulation using a synthetic EEG generated with a real response (for ISI in the range 480-960 ms), contaminated with pink noise.

404 B. Experimental results with simulations

405 Figure 3 shows the effect of the latency-dependent low-pass filtering in the experiments
 406 involving simulations. The plots represent the AEP responses: reference \mathbf{x}_{ref} , not filtered
 407 \mathbf{x} , and latency-dependent filtered \mathbf{x}_{lp} with resolutions of 200, 80, 40 and 10 samples/decade.
 408 Compared with the clean reference response, the not filtered response is affected by noise
 409 due to the noise added to the EEG. As observed, the latency-dependent filtering improves
 410 the quality of the responses by removing the high frequency noise. The noise reduction is
 411 more effective as the latency-dependent filtering is more restrictive. The last plot (for 10
 412 samples/decade) provides the most effective noise reduction, but this filtering is excessive and
 413 causes an important distortion in the AEP response. The resolution of 40 samples/decade
 414 provides the best balance between noise reduction and distortion. The representation of the

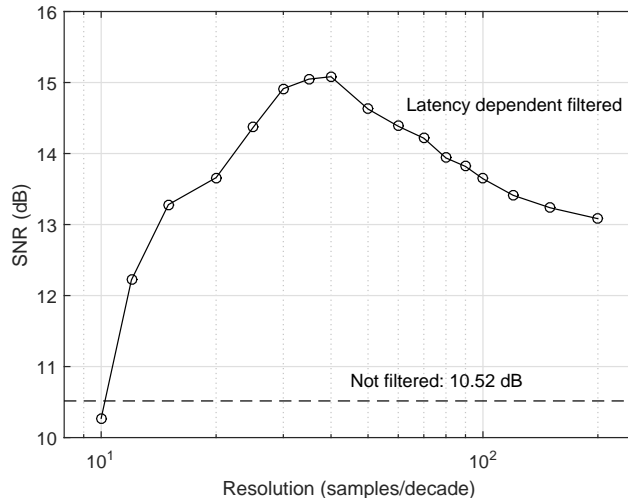


FIG. 4. SNR of the AEP responses as a function of the resolution K_{dec} used in the latency-dependent filtering (solid line with circles). The dashed line represents the SNR for the not filtered response. Results corresponding to simulations; SNR evaluated using \mathbf{x}_{ref} as reference.

415 not filtered AEP response includes 14 700 samples. In the case of resolutions of 200, 80,
 416 40 and 10 samples/decade, representing the AEP response requires 446, 210, 117 and 35
 417 samples, respectively, which implies a substantial reduction of the dimensionality.

418 The noise reduction provided by the latency-dependent filtering has been evaluated in
 419 terms of the SNR. Figure 4 represents the SNR (using the AEP response \mathbf{x}_{ref} as reference) as
 420 a function of the resolution applied in the latency-dependent filtering (solid line with circles).
 421 The dashed line is the SNR for the not filtered response (10.52 dB). As observed, the latency-
 422 dependent filtering improves the quality by appropriately removing the high frequency noise.
 423 The noise reduction is more effective as the filtering is more restrictive, increasing from
 424 13.08 dB (at 200 samples/decade) to around 15 dB (at 40, 35 and 30 samples/decade).
 425 As expected, the resolution providing the best results is around 40 samples/decade (which
 426 is the resolution used for preparing the reference AEP response). A resolution below 30

427 samples/decade reduces the SNR due to the distortion caused for so restrictive latency-
 428 dependent filtering.

429 Section 11 of the supplementary materials¹ contains more detailed results involving sim-
 430 ulations. Figures similar to 3 and 4 are provided for simulations using white noise and
 431 real EEG noise. The behavior with real noise is similar to that with pink noise, with the
 432 best performance around 30-40 samples/decade and improvements greater than 4 dB (sug-
 433 gesting pink noise as a reasonable model for EEG contamination). The improvement is
 434 more important in the case of white noise (around 20 dB), because high frequency noise is
 435 more aggressive in this case (even though this is not a realistic noise model for EEGs). As
 436 observed, thanks to the latency-dependent filtering and down-sampling, as the resolution
 437 decreases (from 200 to around 40 samples/decade), both the SNR and the dimensionality
 438 reduction improve. Beyond 40 samples/decade the latency-dependent filtering is excessive
 439 and produces some distortion in the AEP waves.

440 C. Effect of the latency-dependent filtering with real responses

441 Taking into account the expected spectral content of the AEP responses, the analysis of
 442 table I and the responses estimated in the simulations, a resolution $K_{dec}=40$ samples/decade
 443 has been selected for filtering the AEP responses in the experiments with real recordings.
 444 Figure 5 shows the AEP responses for subject 1. The results are represented for different
 445 ISI configurations from 480-960 ms (top) to 15-30 ms (bottom). Three plots are shown for
 446 each configuration (for consistency evaluation), each one corresponding to the estimation
 447 from an EEG portion of 228 s. The plots in the left panel correspond to the not filtered

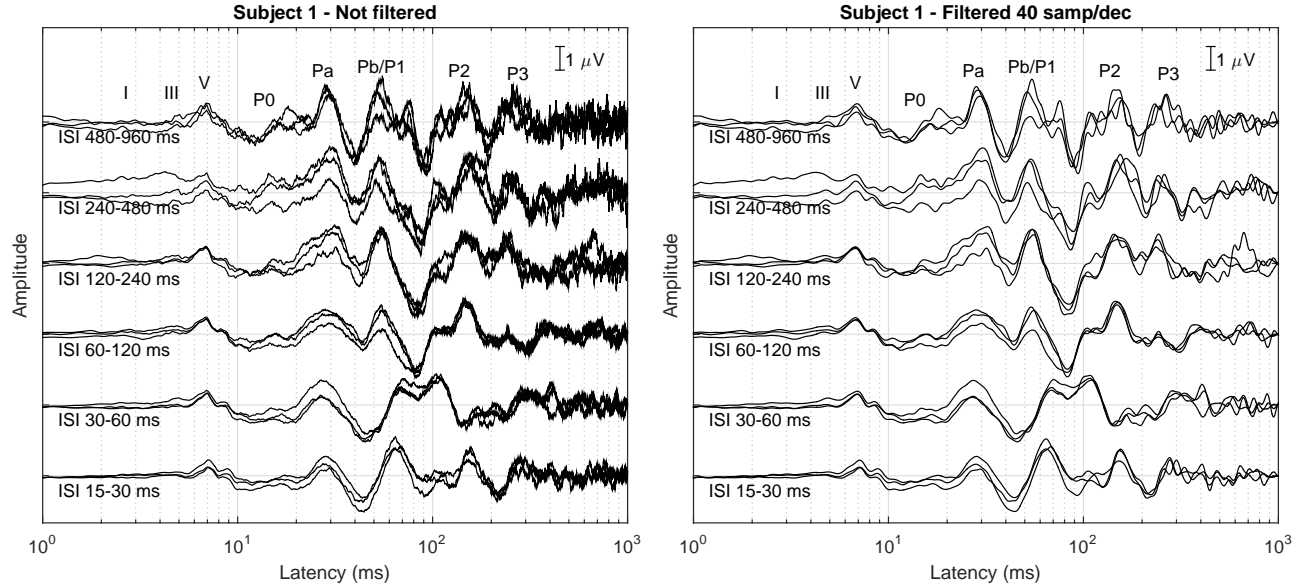


FIG. 5. AEP responses for Subject 1 without and with latency-dependent low-pass filtering (left and right panels, respectively).

448 AEP responses, while those in the right panel correspond to the latency-dependent filtered
 449 ones. At early latency, the plots without and with filtering are similar. However, at late
 450 latency, the not filtered responses are strongly affected by noise and the latency-dependent
 451 filtering provides an effective noise reduction with an evident quality improvement. Section
 452 12.1 of the supplementary materials¹ includes similar plots for the eight subjects included
 453 in this study. The results in figure 5 are consistent with those in the simulations and also
 454 with those for the rest of subjects.

455 Figure 6 shows the SNR results with real AEP responses, for both the grand-average-
 456 based (averaged for the six stimulation conditions) and the subject-based (averaged for the
 457 six stimulation conditions and the eight subjects) estimations, as a function of the resolution.
 458 In the case of subject-based estimations, the SNR progressively increases as the resolution
 459 decreases, from 7.54 dB (not filtered responses) up to 10.74 dB (for resolution $K_{dec}=15$

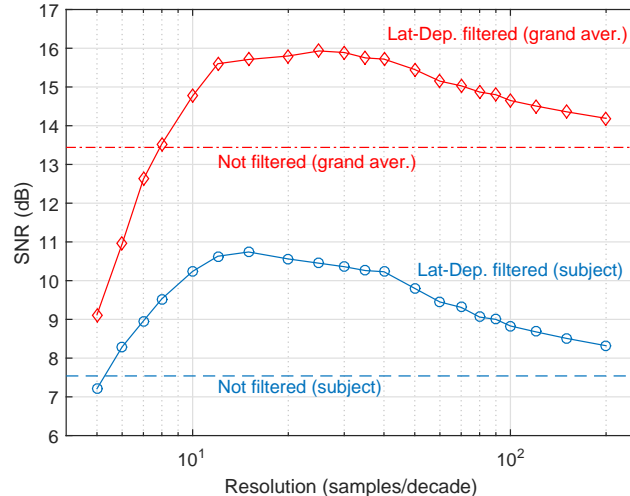


FIG. 6. (Color online) SNR of the real AEP responses as a function of the resolution K_{dec} used in the latency-dependent filtering, for both grand-average-based and subject-based estimations. The dashed lines represent the SNR for the not-filtered responses.

460 samples/decade) because the high-frequency noise is more effectively reduced by a more
 461 restrictive latency-dependent low-pass filtering. Below this resolution, the SNR decreases
 462 due to the distortion caused by the application of a too restrictive low-pass filtering. In the
 463 case of the grand-average-based estimations, the SNR is higher (for both the not-filtered and
 464 the latency-dependent filtered responses) and the maximum SNR (15.93 dB) is achieved at
 465 $K_{dec}=25$ samples/decade. Since the noise level is smaller in the grand-average responses,
 466 the highest SNR is achieved at a resolution slightly greater than in the previous case, and
 467 as expected, the best resolution depends on the noise level. According to this analysis, the
 468 most appropriate resolution depends on the noise level and would be in the range between
 469 15 and 40 samples/decade, requiring 50 and 117 samples in the reduced representation,
 470 respectively. The supplementary materials¹ (Sections 12.2 and 12.3) include detailed results
 471 of the latency-dependent filtered AEP responses and SNRs at different resolutions.

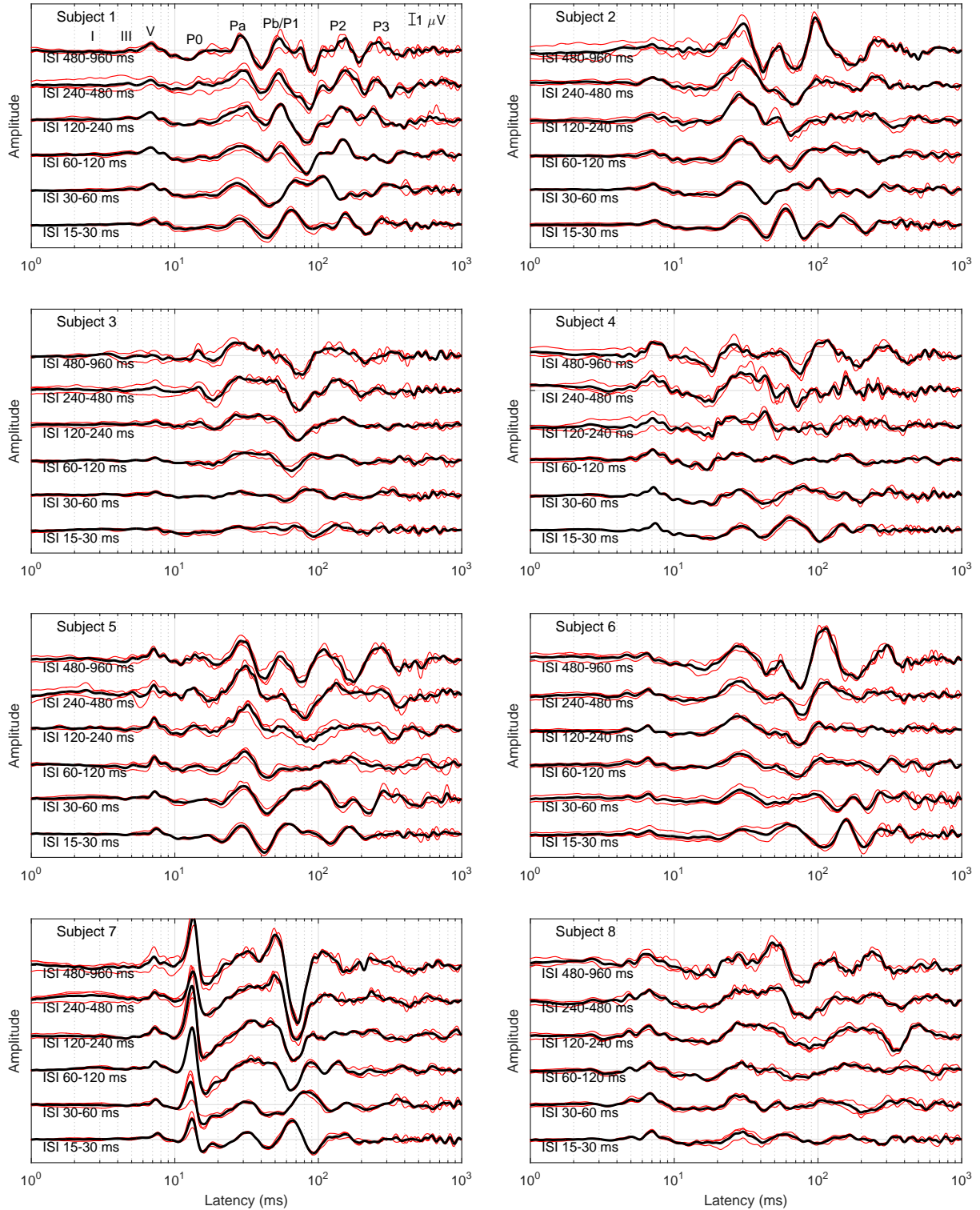


FIG. 7. (Color online) AEP responses estimated for all the subjects with latency-dependent filtering. Filtering configured with 40 samples/decade and responses reconstructed with 200 samples/decade between 1 and 1000 ms. Thin lines are the individual AEP responses estimated from 228 s EEG portions; thick lines are the average AEP responses from the three EEG portions (684 s) for each subject and stimulation condition.

472 **D. Representation of the complete auditory pathway response**

473 The compact representation with 117 samples (for $K_{dec}=40$ samples/decade) for each
 474 AEP response is appropriate for data storage or for advanced data processing procedures
 475 (for instance classification or parameterization of responses (Valderrama *et al.*, 2014c)),
 476 since it minimizes the redundancy without relevant information loss. However, the compact
 477 representation $\mathbf{x}_r = V_r \cdot \mathbf{x}$ is not appropriate for a visual inspection by an audiologist or
 478 for comparison with conventional AEP responses, because the orthonormalization produces
 479 a latency-dependent alteration of the amplitude (as discussed in section II G). The recon-
 480 structed version (in the original representation) is easily obtained as $\mathbf{x}_{lp} = V_r^T \cdot \mathbf{x}_r$, but this
 481 representation is extremely redundant (14 700 samples required for each AEP response). In
 482 order to minimize this redundancy and at the same time provide a representation appro-
 483 priate for an audiological analysis, we have reconstructed the responses with a resolution
 484 of 200 samples/decade in the interval 1-1 000 ms (i.e., three decades), therefore obtaining a
 485 representation requiring 600 samples.

486 Figure 7 represents the AEP responses, for the eight subjects included in this study,
 487 latency-dependent filtered with resolution of 40 samples/decade and reconstructed in the
 488 interval 1-1 000 ms with 200 samples/decade. For each stimulation condition, three responses
 489 are shown (estimated from each 228 s EEG portion) as well as the average (from the whole
 490 684 s EEG) in order to allow the evaluation of the responses consistency. The most relevant
 491 waves are marked in the plot for subject 1. As observed in this figure, most of the AEP waves
 492 (including ABR, MLR and CAEP components) are consistently identified in all subjects.

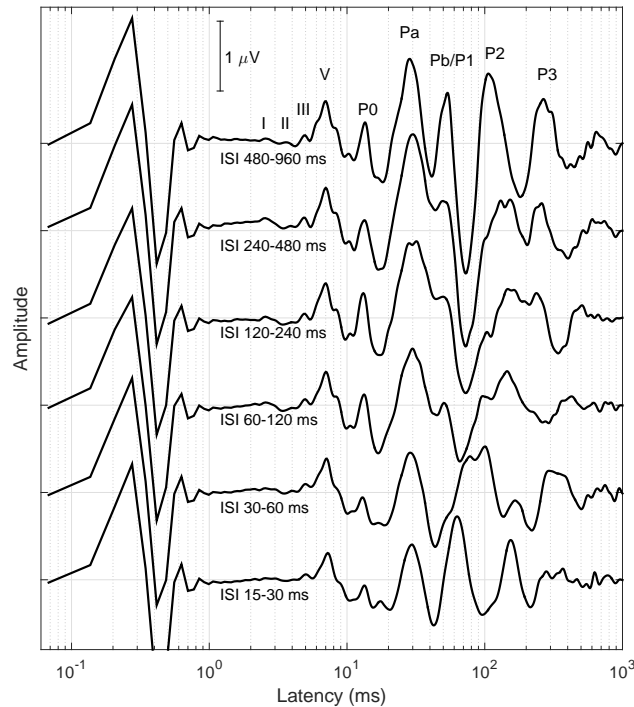


FIG. 8. Grand average of the AEP responses filtered with a resolution of 40 samples/decade reconstructed in the interval between 60 μ s and 1 000 ms.

493 Additionally, some changes in the AEP response morphology associated to the stimulation
 494 rate are appreciated, particularly for the MLR and CAEP components. These changes are
 495 consistent across subjects, and reflect how both peripheral and central structures of the
 496 ascending auditory pathway respond to different acoustic scenarios. This figure also shows
 497 that the responses obtained from subject 7 are affected by the post-auricular muscle (PAM)
 498 artifact, a strong component of myogenic origin that appears at around 15 ms from stimulus
 499 onset (Pratt, 2007).

500 Figure 8 provides the grand-average of the AEP responses obtained from the eight sub-
 501 jects filtered with a resolution of 40 samples/decade and reconstructed in the interval be-
 502 tween 60 μ s and 1 000 ms (in order to provide, in addition to the evoked response, the

503 stimulus artifact and some pre-stimulus response). According to the synchronization con-
 504 figuration, the AEP waves in this figure (as in the rest of figures) are affected by a group
 505 delay of about 1 ms. The stimulation artifact in the interval 100-600 μs can be observed
 506 before the ABR waves. Since the different stimulation conditions only differs in the stim-
 507 ulation rate (but not in the stimulation level) the artifact is similar for all the conditions.
 508 This figure clearly shows changes in the morphology of the MLR and CAEP components
 509 associated to the stimulation rate. The supplementary materials¹ (Section 13) compares the
 510 grand-average responses with and without the latency-dependent filtering.

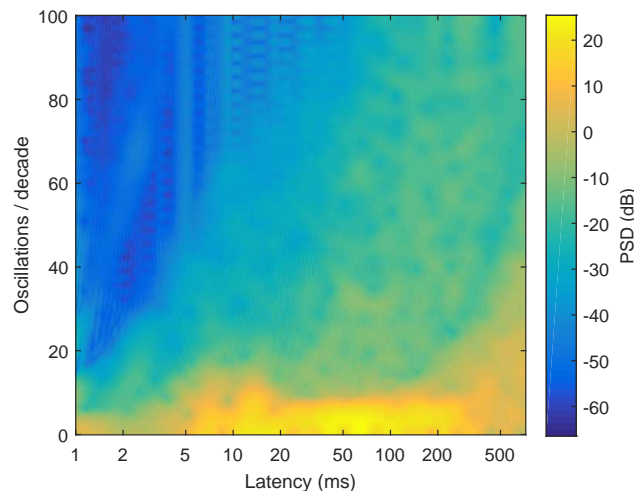


FIG. 9. (Color online) Average spectrogram of the log-scaled-latency AEP responses: power spectral density as a function of the number of oscillations per decade (vertical axis) and the log-scaled latency (horizontal axis).

511 E. Spectral distribution of the energy in the AEP responses

512 The expected morphology of the AEP responses suggests a limit for the number of oscilla-
 513 tions per decade associated to the response waves. On the other hand, the latency-dependent

514 low-pass filtering and down-sampling with 40 samples/decade have provided consistent AEP
 515 responses. In order to investigate the spectral distribution of the energy in the AEP re-
 516 sponses, we propose a spectral analysis with a logarithmic compression of the latency axis,
 517 which allows the estimation of the power spectral density (PSD) as a function of the number
 518 of oscillations per decade. We have estimated the spectrogram of the AEP responses filtered
 519 with $K_{dec}=200$ samples/decade and reconstructed with 200 samples/decade in the interval
 520 0.6-1 000 ms. Figure 9 shows the spectrogram (resulting from averaging the spectrograms
 521 for the eight subjects and the six stimulation conditions). The colormap represents the PSD
 522 as a function of the latency (log-scaled, in the horizontal axis) and the number of oscilla-
 523 tions per decade (in the vertical axis). In this diagram, the frequency can be estimated as
 524 $f(t, f_{dec}) = f_{dec}/(t \cdot \ln(10))$, where t is the latency and f_{dec} is the frequency expressed in
 525 oscillations per decade.

526 As observed in the average spectrogram, most of the energy is below 15 oscillations/decade
 527 (which is consistent with the spectral content expected in the evoked potentials at different
 528 latencies, (Burkard and Don, 2007; Martin *et al.*, 2007; Pratt, 2007)). This support the use
 529 of resolutions around 30 or 40 samples/decade in the latency-dependent low-pass filtering
 530 and down-sampling procedure. Additionally, this figure shows that AEP responses (which
 531 are not stationary processes since the spectral content strongly depends on the latency) can
 532 be considered a quasi-stationary process in the interval 2-300 ms when represented as a
 533 function of the compressed latency axis. This supports the latency-dependent filtering and
 534 down-sampling procedure proposed in this article. The energy above 10 oscillations/decade
 535 observed for latency greater than 300 ms correspond to brain waves (not synchronized with

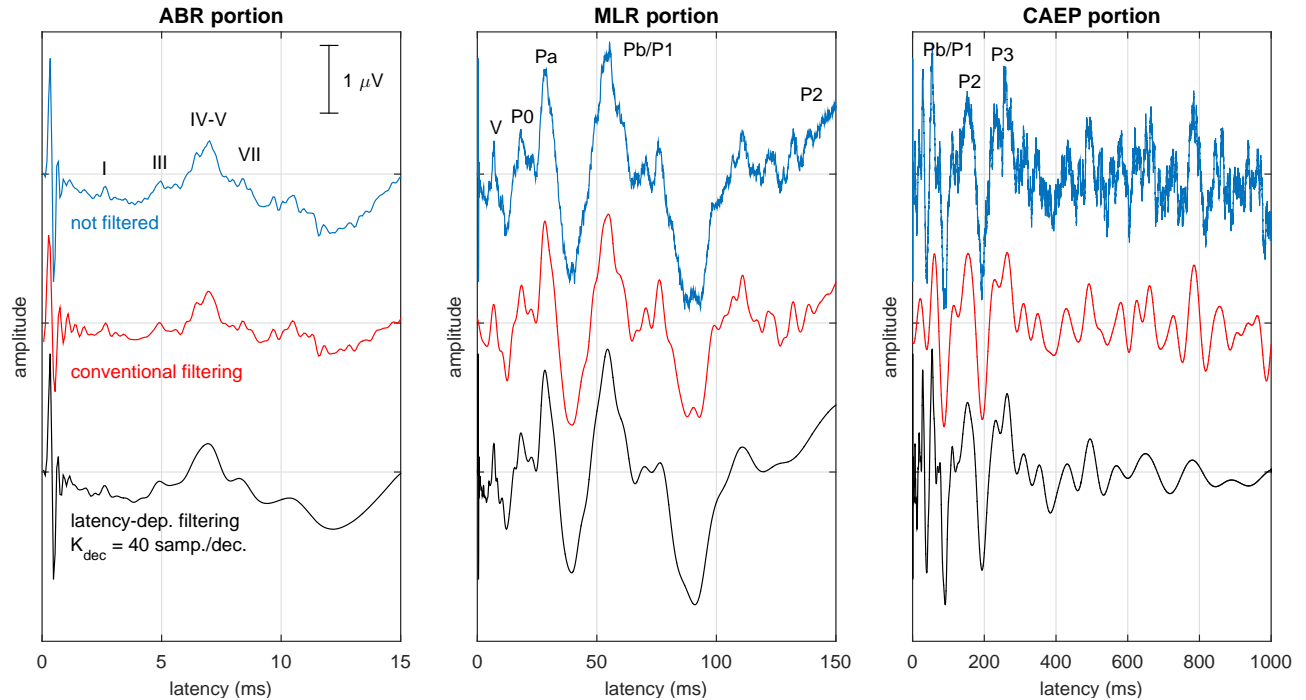


FIG. 10. (Color online) Conventional representation of the AEP responses for ABR, MLR and CAEP components. Responses from a 228 s EEG portion from subject 1, with stimulation configured for ISI=480-960 ms. In each figure, the not-filtered response (top) the response filtered with conventional filtering (middle) and the latency-dependent filtered response (bottom) are compared.

536 auditory stimulation, and therefore noise of neural origin for the AEP responses). Taking
 537 into account the frequency content of this noise, this activity probably corresponds to alpha
 538 or beta brain waves (the band 10-30 oscillations/decade at 500 ms latency corresponds to
 539 the band 8.6-26 Hz). The portion with the lowest energy (below 40 dB) at early latency
 540 and high frequency is associated to frequency components above 7350 Hz (i.e., half of the
 541 original sampling rate). Section 14 of the supplementary materials¹ provides a more complete
 542 analysis of the spectral distribution of the AEP responses, including the average PSD as a
 543 function of the number of oscillations per decade.

544 F. Comparison with the conventional filtering of AEP components

545 The proposed latency dependent filtering was compared with conventional filtering ap-
 546 plied to the ABR, MLR and CAEP portions of the AEP responses. In order to compare
 547 both, each portion was represented in the corresponding latency range, with a linearly scaled
 548 latency axis. In this comparison, the conventional filtering was implemented with band-pass
 549 zero-phase FIR filters with bandwidths 100-3000 Hz for the ABR portion, 10-300 Hz for the
 550 MLR portion and 1-30 Hz for the CAEP portion. The latency-dependent low-pass filtering
 551 was configured for a resolution $K_{dec}=40$ samples/decade. Figure 10 compares the conven-
 552 tional and the latency dependent filtering for the ABR, MLR and CAEP portions. The
 553 not-filtered responses are also included as reference. This response corresponds to the AEP
 554 estimation from a 228 s EEG portion recorded from subject 1 at ISI configuration 480-960
 555 ms. Since the response was estimated from a relatively short EEG portion, it is strongly
 556 affected by noise, and the effect of the noise is better observed. Similar figures for a response
 557 estimated from a longer EEG (684 s) and grand average responses from the eight subjects
 558 (less affected by the noise) can be found in the supplementary materials¹ (Section 15).

559 As can be appreciated in this figure, both the conventional and the latency-dependent
 560 filtering provide an effective reduction of the high frequency noise, and a good synchroniza-
 561 tion of the waves (there is no latency distortion since both methods apply zero-phase filters).
 562 However, there are differences between the filtered responses provided by both methods. The
 563 conventional filtering is too restrictive at the early portion of the responses, and too per-
 564 missive in the late portion. This produces a distortion of the ABR components in the MLR

565 plot, and of the MLR components in the CAEP plot, while noise is insufficiently attenuated
566 at the late portions of the responses. The better preservation of the wave components and
567 the more effective noise reduction provided by the latency-dependent filtering is associated
568 to the continuous variation of the bandwidth with the latency. Additionally, the proposed
569 method avoids the discontinuity imposed to the conventional AEP analysis, allowing a more
570 comprehensive interpretation of the responses of the complete auditory pathway (as can be
571 observed in figures 7 and 8).

572 There are also slight differences in the low frequency components of the responses filtered
573 by both methods. Since the conventional filter applies band-pass filtering (but the proposed
574 latency-dependent filtering is low-pass) some low frequency components are attenuated in the
575 former but are not in the latter (this is evident, for example, in the last portion of the ABR
576 response). High-pass filtering is particularly necessary when an isolated portion (for example
577 ABR) is estimated with stimulation at high rate, in order to avoid the interference from late
578 components elicited from the adjacent stimuli. However, in the experiments included in
579 the present study, this interference is minimized since the whole AEP response (including
580 ABR, MLR and CAEP components) is modeled and a deconvolution procedure (rather than
581 a simple average) is applied for the response estimation, and the estimations provided by
582 the latency-dependent low-pass filtering are appropriate. In any case, the proposed method
583 could be easily adapted in order to provide band-pass filtering instead of low-pass filtering.

584 **IV. DISCUSSION AND CONCLUSIONS**

585 In this article we present a procedure providing latency-dependent low-pass filtering and
586 down-sampling to be applied to AEP responses from the complete auditory pathway. The
587 procedure is formulated as a matrix operation. An orthonormal matrix V_r applied to the
588 original AEP response provides its projection in the subspace of the latency-dependent
589 band-limited functions, i.e., a compact representation of the filtered signal. The compact
590 representation can be transformed to the original representation (using the transposed V_r^T
591 matrix) or, alternatively, the response can be reconstructed at a specific set of latencies.

592 The latency-dependent filtering and down-sampling is implemented by applying a uniform
593 filtering and down-sampling in the compressed latency axis, using root-raised cosine sampling
594 functions. A linear-logarithmic compression (with a resolution specified in terms of the
595 number of samples per decade) has been applied.

596 The proposed procedure has been evaluated with both simulations and real AEP re-
597 sponses. In the experiments presented in this paper the dimensionality has been reduced
598 from 14 700 samples (in the original representation) to 117 samples (in the compact represen-
599 tation for a resolution of 40 samples/decade). The procedure provides a significant quality
600 improvement of the AEP responses, associated to the reduction of the high frequency noise
601 at late latency. This improvement is clearly observed in the appearance of the responses and
602 was objectively measured in terms of the SNR. For resolutions from 200 to 40 or 30 sam-
603 ples/decade, the dimensionality reduction is accompanied of quality improvement. Below 30

604 or 20 samples/decade (depending on the noise affecting the responses) the dimensionality
605 can be reduced but the quality degrades due to the distortion of the waves.

606 The proposed method allows an adequate filtering and representation of the complete
607 auditory pathway response. When the auditory response is estimated separately (i.e., when
608 ABR, MLR or CAEP are independently measured), the conventional filtering and the uni-
609 form sampling are appropriate (because each portion is contained in just one decade of
610 latency). However, conventional filtering and sampling are not appropriate for the response
611 of the complete auditory pathway (covering almost three decades) because of the noise
612 effect and the redundant representation. The alternative of representing each portion of
613 the auditory response separately creates an artificial discontinuity between the waves in
614 the different portions, not appropriate for global analysis and interpretation of the evoked
615 responses. The latency-dependent filtering and down-sampling locally approaches the filter-
616 ing and sampling-rate conventionally applied for ABR, MLR and CAEP, but at the same
617 time eliminates the discontinuity between the different portions. This way, the proposed
618 procedure offers new perspectives for the design of audiological experiments and for the
619 analysis of evoked responses, allowing the simultaneous study of all the waves generated
620 by the complete auditory pathway. While conventional filtering applies band-pass filtering,
621 the proposed latency-dependent filtering applies a low-pass filtering. This is not a strong
622 limitation for the representation of the complete auditory pathway, since the deconvolution
623 applied to obtain the AEP responses avoids the interference of the late components over
624 the early components. However the proposed latency-dependent filtering can be adapted
625 for band-pass filtering, of potential utility for other audiological experiments. The proposed

626 filter design with zero-phase RRC filters guarantees that the latencies of the waves are not
627 delayed by the procedure, but the non-causality associated zero-phase filters should be taken
628 into consideration, depending on the purpose of the AEP analysis (de Cheveigne and Nelken,
629 2019).

630 This manuscript includes a study of the spectral content of the AEP responses. From the
631 responses represented in the log-scaled latency axis, the AEP responses can be considered
632 a quasi-stationary process, at least in the interval 2-300 ms. This quasi-stationarity of the
633 responses has allowed the effective reduction of the noise without distortion by applying
634 the latency-dependent filtering. The spectral analysis with the log-scaled latency axis also
635 reveals that the spectral range below 7 oscillations/decade accumulates 95% of the AEP
636 energy (see the supplementary materials ¹, Section 14). This suggests that in the case
637 of responses severely affected by noise, a more aggressive latency-dependent filtering (for
638 instance with 20 or 15 samples/decade) would be useful (since in spite of the distortion of
639 some waves, most of the shape of the AEP would be preserved).

640 In this article, the compression of the latency axis was performed with the relatively
641 simple linear-logarithmic compression. More flexible compression functions could be applied
642 in order to obtain a more accurate control of the local sampling frequency for each latency,
643 or even for including pre-stimulus negative latency also with compression (i.e. including,
644 with appropriate latency compression, portions where the response is expected to be null),
645 in order to help the audiologist in the verification of the AEP response consistency, or in
646 order to allow an evaluation of the SNR based on the pre-stimuli response (Polonenko and
647 Maddox, 2019).

648 The advantages of the proposed procedure are obtained with a minimum computational
649 cost, since only a matrix product is required. Additionally, the procedure provides a com-
650 pact representation of the AEP responses (i.e., with the minimum number of samples and
651 without information loss), which reduces the requirements for storage or transmission of
652 AEP databases (of potential utility, for example, for remote AEP recording or monitoring).
653 It can be also applied for reducing the computational cost of deconvolution algorithms for
654 AEP responses (for example the IRSA algorithm could be equivalently applied in the orig-
655 inal or in the reduced representation space, with a substantial reduction of the memory
656 requirements and execution time in the reduced representation space). On the other hand,
657 the concentration of the relevant information in a reduced number of samples simplifies
658 the post-processing of AEP data. Algorithms for classification, characterization or param-
659 eterization of waves or AEP responses (Bradley and Wilson, 2005; Fridman *et al.*, 1982;
660 Kamerer *et al.*, 2020; Valderrama *et al.*, 2014c), as well as procedures based on artificial
661 intelligence, involving deep artificial neural networks or designed under the perspective of
662 big-data analysis (Dobrowolski *et al.*, 2016; Mosqueda-Cárdenas *et al.*, 2019) would benefit
663 from compacting the relevant information in low-dimensionality vectors (Trunk, 1979).

664 The procedures for latency-dependent low-pass filtering and down-sampling and for the
665 response reconstruction have been implemented as MatLab/Octave functions and are pro-
666 vided in the supplementary materials¹ (Sections 6 and 9, respectively). A MatLab/Octave
667 script has also been prepared for running a demonstration involving the proposed proce-
668 dures (see the supplementary material¹, Section 16). The script reads an AEP response,
669 estimates the latency-dependent filtering and down-sampling matrix and the reconstruction

670 matrix for the specified resolutions, and represents the not filtered responses as well as the
671 filtered responses (a) in the reduced representation, (b) in the original representation, and
672 (c) at the latencies specified for reconstruction. The script also plots the functions of the
673 basis. Additionally, in order to provide the community with these computational tools,
674 MatLab/Octave functions and scripts as well as data with examples have been included in
675 a compressed directory in the supplementary materials¹.

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680 ¹See the supplementary materials at [URL will be inserted by AIP] for a PDF file presenting (section 1) a
681 description of the matrix implementation of low-pass filtering and down-sampling; (section 2) a description
682 of the matrix implementation of the latency-dependent low-pass filtering and down-sampling; (section 3)
683 the derivation of the linear-logarithmic compression of the latency axis; (section 4) a description of the root-
684 raised cosine filters; (section 5) a description of the latency-dependent filtering and down-sampling using
685 RRC filters; (section 6) MatLab/Octave code providing the latency-dependent filtering and down-sampling
686 matrix; (section 7) a description of the functions in the orthonormal basis; (section 8) a description of
687 the procedure for response reconstruction at a specified set of latencies; (section 9) MatLab/Octave code
688 providing the reconstruction matrix; (section 10) a description of the response used for the simulations; (sec-
689 tion 11) detailed experimental results with simulations; (section 12) detailed experimental results with real
690 responses; (section 13) grand average of the responses of the auditory pathway; (section 14) spectral analysis
691 of the AEP responses; (section 15) MatLab/Octave code of a script for testing the latency-dependent filter-
692 ing and down-sampling. A compressed directory with examples and MatLab/Octave scripts and functions,
693 aiming to help the reader apply the latency-dependent low-pass filtering and down-sampling procedure
694 described in this paper is also included in the supplementary materials.

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